

Simulation and designing of Voltage Controlled Oscillator by using C-MOS technology: A review

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Abstract-

“CMOS” refers to digital circuit design, and the processes used to implement that design on integrated circuits. CMOS circuits in VLSI consume less power when static. This paper presents a brief study of various types voltage-controlled oscillator and their function like a complementary metal–oxide–semiconductor (CMOS) ring oscillators and LC-VCO.

The high performance VCO on 45 nm technology to achieve the desired objectives such as both non-linear and linear operations. The circuits used is a modified design of VCO.

The study of an improved design of four-stage CMOS differential ring voltage-controlled oscillator(VCO)withhigh-outputfrequency,lowphasenoise, andlowpowerconsumptionis proposed in this paper. A new differential delay cell has been used for differential ring VCO whichutilising dual-delay-path topologytoattainbothhigh-outputfrequencyandlowphase noise. The simulation results have been obtained in TSMC 0.18- μ m CMOS process with a supply voltage (Vdd) 1.8 V. The proposed design of VCO exhibits an output oscillation frequency range from 1.619 to 3.712 GHz. The power consumption for this frequency range variesfrom4.628to10.545mWwithcontrolvoltagevariationfrom0.1to1.0V.Theproposed designoccupiescompactlayout areaof0.207 mm² andachieves –89 dBc/Hzphasenoiseat 1 MHz offset from 3.712-GHz carrier frequency. Improved performance for this design has beenachievedintermsofoutputoscillationfrequency,phasenoise,andpowerconsumption.

Keywords: Voltage control oscillator, ring oscillator, low power consumption, cmos- technology, phase noise, oscillation frequency etc.

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I. Introduction:

Electronicoscillatorsbelongtotheveryfirstelectriccircuitsatthebeginningof thetwentiethcenturyacompletesystematicdesignconceptforthisclassofelectroniccircuitsisnotavailableuntil now.Oneofthereasonsisthat thebehaviorofthesedynamicalcircuits depends in an intrinsic manner on the nonlinearities of within the circuit and therefore we areconfrontedwithnonlineardifferentialequations.The eoscllatoryrcircuitbehaviorisrelated fromamathematicalpointofviewtotheso-calledlimitcycles.However electronicoscillators arefascinatingcircuitsinmanysensebecause theprogres sinmanufacturingtechnologiesof electronic devices and circuits led to new challenges in oscillator modeling and new mathematical concepts for

solving the descriptive equations of oscillators. In certain cases the behavior of an electronic oscillator should be influenced by the behavior of other electronic systemsina desiredmannersuchthat entrainmentandsynchronizationeffects

arise. Therefore driven nonlinear oscillators and their descriptive equations have to be consideredwhereevenchaoticbehaviorcanappear.From aphysicalpointofviewelectronic oscillators can be interpreted as such systems where dissipative structures occur. Ilya Prigogine coined this phrase as a name for the patterns which self-organize in far-from- equilibrium dissipative systems and limit cycles are a special case of them; see Nicolis, Prigogine [1]. Such dissipative systems are nonlinear and have to be connected with a DC

power supply for delivering energy into the system. Furthermore, these systems interact with a heat bath where energy is dissipated.

Accordingly, the fluctuations of the heat bath influence electronic oscillators. Electronic oscillators belong to the very first electric circuits at the beginning of the twentieth century. A complete systematic design concept for this class of electronic circuits is not available until now. One of the reasons is that the behavior of these dynamical circuits depends in an intrinsic manner on the nonlinearities of the circuit and therefore we are confronted with nonlinear differential equations. The oscillatory circuit behavior is related from a mathematical point of view to the so-called limit cycles. However, electronic oscillators are fascinating circuits in many senses because the progress in manufacturing technologies of electronic devices and circuits led to new challenges in oscillator modeling and new mathematical concepts for solving the descriptive equations of the oscillator as electronic noise. As a result, electronic oscillators have to be modeled by driven nonlinear stochastic differential equations with limit cycle type solutions. In most cases, analytical solutions for this type of equations are not available and approximation concepts have to be developed. From this point of view, electronic oscillator circuits are until now a source of inspiration for new mathematical and physical concepts; see e.g. Guckenheimer [2], Fig. (1). Dissipative structure.

However, for the oscillator circuit design, not only approximative solutions of certain descriptive equations are needed since at the first stages of a design process only very few circuit parameters are known. Note that a circuit design concept consists of two steps where it is starting from the specifications of a circuit under design. These specifications are closely related to the solutions of the descriptive equations of the designed circuit. In a first design step, the circuit architecture - circuit shape (O'Dell [3]) - has to be chosen, whereas in the second step, the free parameters of this circuit shape have to be determined. Therefore, if not all circuit parameters are available, it is even not possible to know whether the descriptive equations possess oscillatory or limit cycle solutions.

For this reason, Mandelstam and Papalexi [4] developed in 1931 the concept of parametrized descriptive equations for oscillator circuits based on ideas of the French mathematician Henry Poincaré. This concept was also the basis for the bifurcation theory of electronic oscillators; today it is called the Andronov-Hopf theorem in the theory of dynamical systems. In contrast to the quantitative analysis of nonlinear differential equations, this theorem studies these equations from a qualitative point of view. By

means of the qualitative analysis, we are able to consider the qualitative change of different types of solutions of nonlinear differential equations in dependence on certain circuit parameters. In the case of the Andronov-Hopf

theorem, the change from an equilibrium point to a limit cycle is explained. Although this theorem is known in electronic oscillator analysis since 1935 [5] (see Maggio et al. [6] for a more recent publication), it was never used for a systematic design process of electronic oscillators until recently; see Mathis & Russer [7], Prochaska et al. [8]. As mentioned above, the circuit description with parametrized equations is well suited for the second step of the design process and therefore the Andronov-Hopf theorem should be used in design processes for electronic oscillator circuits; a first concept idea was presented by Mathis [9].

II. Literature Review

A- Early Oscillator Circuits

Since in 1895 Marconi showed for the first time that the laboratory arrangement of Hertz can be used for the wireless transmission of information, along larger distances more powerful electrical arrangements were desired. Around 1900 several researchers (e.g. Thompson, Tesla, Slaby, Braun, von Arco and others) suggested improved versions of Marconi's arrangement. Probably it was a milestone as Duddell published his paper "On rapid variations in the current through the direct-current arc" [10] where he used results of the German physicist Simon [11]. A few years later Poulsen improved Duddell's oscillatory generator substantially from a technical point of view and as a result he presented in 1906 a new powerful arrangement with an arc as electronic device for wireless transmission of telephony signals. For further studies we refer to Blake [12] and Nesper [13]. Although the physical processes in arcs are rather difficult to understand at this time because of their electronic nature, reasonable nonlinear models were developed by Kaufman, Duddell, Simon, Wagner and others. In the dissertation thesis of Wagner, many aspects of such circuits were discussed [14]. Using these results, Zenneck [15], a former co-worker of the above mentioned Braun, published in 1914 an interesting paper where he studied the start-up behavior of such RLC circuit including a nonlinear arc device. In contrast to his predecessors, he described the behavior of the circuit by means of a nonlinear differential equation that described the energy (or power) balance. After solving this equation, he got the approximative solution that is similar to the approximative solution of the van der Pol equation, which was discovered several years later by van der Pol in the analysis of triode

oscillators. From a mathematical point of view Zenneck's balance equation correspond to an approximative first integral of the van der Pol equation. Zenneck's paper was also the first that studied the start-up behavior of oscillatory circuits in more details. Further details are considered in an other publication [16]. Circuits including sparks or arcs were the first successful generators for oscillatory currents. These electronic circuits had several disadvantages. Although engineers and physicists in the leading industrial companies (e.g. Marconi Comp., Telefunken, AT&T, Western Electric Comp.) tried to improve these circuits by using interesting ingenious ideas the robustness of these arrangements as well as their transmitting power were rather restricted. Therefore a new generation of generators applied and now nonelectronic principle to get high-power oscillatory currents. For this purpose the static frequency doubling effect which was studied by Epstein, Joly and Vallauri was used for the development of rotating alternators (see Kühn [17], Meißner [18]). These generators had much better properties than the spark or arc circuits. Only the frequency of oscillatory waves generated by these electrical machines was restricted. However at this time transmitting stations worked with long waves. Around 1925 the situation was changed as the Heaviside layer was observed by nonprofessional users using transmitting stations with short waves. Within this range of frequencies rotating alternators cannot be applied. Moreover short waves did not need high-power transmitting stations such that the power can be reduced for these frequencies. As a conclusion electronic vacuum tubes were used to build generators with oscillatory behavior for powerful transmitting stations.

B-Tubes and Oscillators

First ideas for a new electronic device were published by Fleming in 1904. For the invention of his thermionic diode (or Fleming detector) he used research results in the area of emission and transport of electrons (Edison effect, Richardson, Wehnelt; see Johannsen [19]) in vacuum although the physical details were not well-known at this time. A modulation of the current in Fleming's thermionic diode was achieved by adding a grid. This was done by de Forest in 1906. This electronic device was called by de Forest as Audion [20]. In the same year v. Lieben presented a patent of triode type of amplifier valve [21] that was improved by him in a patent from 1910 together with Reisz and Strauß [22]. The audion as well as the "Lieben-valve" were three-pole devices with cathode, grid and anode. In contrast to

de Forest's audion the "Lieben-valve" was filled with mercury. Therefore Lieben's valve was called "gas relay" and de Forest's "electron relay" (audion). This difference remained unclear in the following years and led to many discussions; see e.g. Armstrong [23] (p. 220) and Meißner [24] (p. 65).

Further references can be found in a paper of Tucker [25]; see also Johannsen [19]. After the discovery of these triode valves it lasted further six years until first practical circuits were available. The first two classes of circuits were amplifiers and oscillators. Several groups in Germany, USA, United Kingdom, and Austria were participated in these activities and there were many relationships between these groups. Therefore it is difficult to solve the problem of priority with respect to the different electronic circuits; see e.g. Tucker [25], Johannsen [19], Barkhausen [26] (part II, pp. 112), Meißner [24], and Skowronnek [27]. In 1913 Meißner presented a first high frequency transmitter including an electronic oscillator with a von Lieben tube (Fig. 2). It should be remarked that the principles of thermionic conduction were studied based on physical foundations by Langmuir.

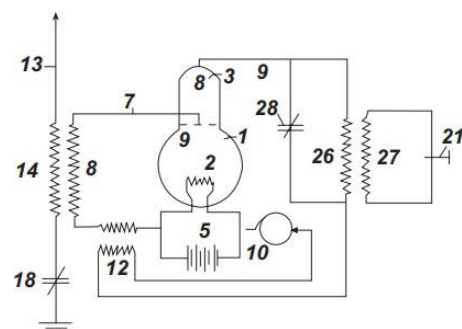


Fig. (2). Meißner's oscillator with a v. Lieben tube.

A first example of an electronic oscillator with a valve which is the main subject of four papers was invented between 1912 and 1913 where the difference between de Forest's audion and the Lieben-Reisz valve is not considered (but see remarks of Meißner [24] (p. 65) and Hazeltine [30] (p. 98)). In agreement with Hazeltine [30] it seems that Meißner and Armstrong had similar ideas at the same time with respect to an oscillating circuit including a valve. The corresponding comments of Tucker [25] are a little bit obscure. Since the problem of the design of electronic circuits and especially oscillators with valves in order to transmit electromagnetic waves and receive them was a main subject in all military laboratories in all industrial states which are involved in the first world war

many information became a secret. At the end of this war several electrical engineers and physicists published their results with a delay of one or more years (see corresponding remarks in the papers of Hazeltine [30], Barkhausen [26], Meißner [24] and others did not have an opportunity to publish it, just like Colpitts or Hartley). Hazeltine [30] (p. 98) gave some more references to interesting collections of oscillating circuits. Obviously at the end of the first world war many different oscillator circuits were known and several authors began to publish their theoretical results about this interesting class of electronic circuits.

C. Descriptive Equations for Electronic Oscillators

Probably the first theoretical paper about electrical oscillators with a valve was published by Vallauri [31] which was published in German some months later. A sinusoidal oscillatory behavior already exists. By means of this approach Vallauri got “the exact determination of the conditions for oscillations in an audio circuit”. After the publication of this paper many other authors presented results that are more or less equivalent. These results were different in modification of the valve model or the decomposition and its interpretation of the linearized

oscillator circuit. We would like to mention only the comprehensive papers of Hazeltine [30], Heising [33] and Barkhausen [26]. The different approaches were compared by Albersheim [34]. Unfortunately it was known already before 1920 that a linear theory cannot be complete for describing all aspects of electronic oscillators. Whereas the condition of oscillatory behavior can be derived by a linear model a nonlinear model is essential to determine the amplitude of these oscillators. This statement was given by Möller [29] in a very similar manner (p. 331). Based on the idea of the feedback principle for the functionality of electronic oscillators this author developed a theory for these circuits that used again the idea of a power balance equation (just like Zenneck in 1914 for the case of oscillator circuits with an arc). For this purpose Möller developed a concept that takes into consideration only the first harmonics of the oscillatory behavior and a nonlinear differential equation for the circuit was not derived. As a result he got the method of the “oscillatory characteristic” (in German “Schwingkennlinie”) that can be interpreted now as a variant of the harmonic balance or averaging method. Some remarks to the history of this method can be found in the monograph of Sanders and Verhulst [35] (pp. 181).

Again other authors presented similar approaches that correspond to the fundamental mathematical problem of nonlinear oscillatory systems or nonlinear differential equations. In contrast to Möller's approach van der Pol [36] derived in his doctorate thesis for the first time a nonlinear differential equation for an oscillatory electrical circuit and especially for an electronic oscillator circuit including a triode valve. Furthermore he was able to apply a special perturbation method that resulted in a solvable nonlinear differential equation. It was a variant of an averaging method (Lagrange's secular perturbation method) that was known to van der Pol from his studies in physics. In his famous paper from 1920 he states that the equation under consideration is closely related to some problems which arise in the analytical treatment of the perturbation of planets by other planets.” The differential equation of van der Pol became an eminent impact to a new mathematical discipline “nonlinear oscillations” and at the end the mathematical theory of dynamical systems. On the other hand we have to emphasize that although interesting from a theoretical point of view van der Pol's equation was not useful in practical situations at his time for two reasons: 1) Graphical differentiations of higher order are required. 2) Only in the simplest cases the perturbation method leads to a solvable differential equation. In other cases only the steady amplitudes can be calculated. Although Möller's method leads to tedious calculations a semi-analytical approach was presented by Joos [37]. This author started with a good analytical approximation of the characteristic curves in form of an arctan-function and as a result he got another kind of oscillatory characteristics that did not include circuit parameters or valve constants. But also this method did not become popular. Between 1920 and 1929 only very few scientific groups tried to develop new impacts to the theory of electrical oscillators. One of the few exceptions were van der Pol and Appleton. They published very interesting papers about the entrainment problem, forced oscillations and on relaxation problems as well as other aspects of nonlinear oscillators.

The main results of these authors are contained in van der Pol's review paper from 1934 [38]. A new era of oscillator theory began as the Russian school of Mandelstam and Papalexi entered this area. Although Papalexi had published a book about “The Theory of Oscillators with Electronic Valves” in 1922 and Mandelstam was already a well-known scientist who was involved also in aspects of wireless transmission of electromagnetic

waves they considered electronic oscillators from a new point of view. In 1929 Andronov who had Mandelstam as his academic teacher published a brief paper [39] where he applied Poincaré's theory of limit cycles to the van der Pol equation. Another paper together with Witt [40] followed in 1930. In a further paper [41] these two authors showed that Poincaré's theory is useful for studying rather difficult aspects of the entrainment effect in nonlinear oscillations. By means of these results it could be shown that the theory of limit cycles of Poincaré was a suitable framework to study problems in nonlinear oscillations. Another important step towards a unified theory of nonlinear oscillations was the introduction of parametrized nonlinear differential equations and again Poincaré's ideas were used. Mandelstam and Papalexi showed [4] that first aspects of a theorem that is now called Andronov-Hopf theorem and describes a bifurcation of a limit cycle from a static equilibrium point. These ideas became a big impact in Russia for the theory of nonlinear oscillations and several very active groups studied many aspects of this problem. The references of the famous book of Andronov, Witt and Chaikin [42] first published in 1937 illustrate this exciting area. In a review paper from 1935 Mandelstam, Papalexi, Andronov, Chaikin, and Witt [5] presented a complete concept of nonlinear oscillations that contains already almost all modern aspects of this theory; see also Mathis [43] and in particular the papers of Bissell [44], Aubin, Dalmedico [45] and Pechenkin [46]. It should be mentioned that at the same time Krylova and Bogoliubov [47] developed the mathematical foundations of a perturbation theory for oscillatory differential equations that is known now as averaging method or harmonic balance. It lasted more than 40 years until this subject was considered by Mees and Chua [48] in 1979. Afterwards Andronov-Hopf bifurcation became an essential subject in the theory of electrical circuits (e.g. Mathis [16]). It should be remarked that in higher dimensional systems ($d > 2$) additional methods are needed.

Example: Design of a 2.4 GHz LC-Tank VCO In this section we show the application of our proposed design concept to a practical oscillator design. The desired specifications are shown in Fig. (5) (a $0.25\mu\text{m}$ RFCMOS technology from IHP (SGB25V) is used). An inductor with a series inductance of $L_s = 11.1\text{nH}$ is chosen for the requested frequency range. Using our proposed model for the voltage dependent varactor capacitance [64] we are able to approximate the varactor dimensions. Fig. (6) shows

an approximation of the frequency characteristic of the VCO in dependency of the varactor width W_v . Our calculations show, that as a first estimation a varactor width of $W_v = 150\mu\text{m}$ is an appropriate choice for the requested frequency range (Fig. 5)

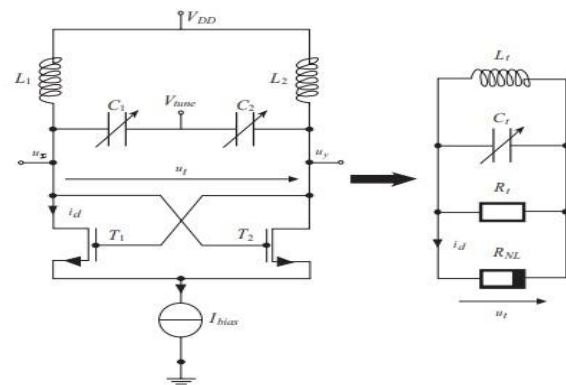


Fig. (4). LC-tank voltage controlled oscillator.

Setting L_v to two times the minimum channel length is a good compromise between series resistance and $C_{v,max}/C_{v,min}$ ratio [65]. In order to minimize the parasitic resistance the length of the transistor pair L_n is set to the minimum channel length. We set the value of I_{bias} to the maximum value allowed by the specifications in order to maximize the output amplitude and optimize the phase noise characteristics of our VCO [66]. All design variables have been determined except the width W_n of the transistor pair. It is guaranteed that a variation of the width of the transistor pair W_n leads to a stable limit cycle as was shown in the previous section. Using expression (7) it is possible to calculate the needed width W_n . We find the bifurcation point and therefore the starting point to fast stable oscillation at a parameter value of $W_n = 2.03\mu\text{m}$. Increasing W_n leads to an expansion of the limit cycle and accordingly to a rising of the oscillation amplitude (8) (Fig. 6).

Higher Dimensional State Space Representation

A higher dimensional state space representation allows the inclusion of other parasitic effects and structural enhancements in comparison to the 2-dimensional state space modeling. Examples of possible parasitic effects could be the nonlinear effects of the cross-coupled transistor pair or substrate effects for instance. An example for a structural enhancement that could be included in the modeling using higher dimensional state space equations is a filtering capacitance parallel to the current source (Fig. 8). In order to carry out the bifurcation analysis of an sinusoidal oscillator

Param.	Spec.	Achieved
f_0	2.4 GHz	2.4 GHz
V_{DD}	2.5 V	2.5 V
I_{bias}	≤ 1 mA	1 mA
TR	≥ 20 %	34 %
v_i	≥ 1 V	≥ 1.37 V
C_{load}	≥ 100 fF	100 fF

Transistor Pair		
Param.	Maple	Cadence
W_n	$> 2.03 \mu\text{m}$	$> 3.05 \mu\text{m}$
L_n	250 nm	250 nm
R_0	10 k Ω	9.84 k Ω
C_{NMOS}	3.97 fF	3.77 fF

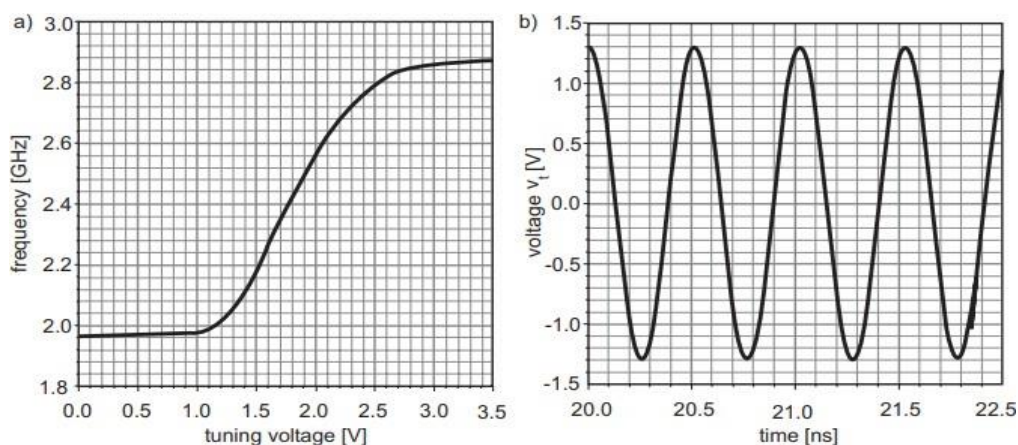
Inductor		
Param.	Maple	Cadence
$L_{i,p}$	11.16 nH	-
$C_{i,p}$	16.6 fF	-
$R_{i,p}$	1824 Ω	-

Varactor		
Param.	Maple	Cadence
W_v	150 μm	140 μm
L_v	500 nm	500 nm
$R_{v,p}$	3222 Ω	-
C_{v0}	378 fF	-

Fig. (5). Specifications and design parameters for $V_{tune} = 0$.

The center manifold approach for simplifying dynamical systems has a nice geometric interpretation. Further investigation of the reduced system with analytical methods are well-known in system theory [2]. The

following methods are predestinated for the use in common with center manifolds because they are formulated only in the 2- dimensional case or easier to handle in that dimension



III. CONCLUSION:

The essential steps of a systematic concept for the design of electronic oscillator circuits. It can be shown that by means of advanced mathematical methods the intrinsic nonlinear problem of oscillator design is manageable. Using a high-efficient computer algebra system (e. g. MAPLE or MATHEMATICA) semi-analytical expressions for the oscillator design can be derived.

Based on previous work further research will be done in order to develop a computer aided design for electronic oscillators in the GHz area. It is our goal to enable expert analog designer to design oscillators by means of a variety of well-adapted tools instead of a general purpose circuit simulator.

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