#### RESEARCH ARTICLE

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# Power Quality Analysis of Harmonics in 3.5kVA Single Phase Generator using Electronic Filter in Matlab/Simulink Environment.

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#### **ABSTRACT**

The aim of this workis to investigate the harmonic content of nonlinear loads connected to 3.5 kVA power supply source and to mitigate the effects using electronic filter (double tuned low pass active shunt filter). The presence of harmonics was investigated by collecting data from the power source, 3.5 kVA using 'Owon Oscilloscope' metering device. The data was then processed by means of Fourier analysis into harmonic spectrums. Mitigate the harmonics found in this power source a double tuned low-pass active shunt filter was design from discrete elements (resistors, capacitors and inductors) and implemented in the MATLAB/SIMULINK environment. The result obtained for  $V_{THD}$  before and after the simulation procedure using the series active passive filter was higher than the expected IEEE standard of 5%; while the result obtained using the double tuned low pass active shunt filter then after simulation gave a satisfactory result less than the expected IEEE standard of 5% after simulation.

#### I. INTRODUCTION:

Electrical AC supply is ideally a pure sine-wave of both 50 or 60 Hz fundamental frequency and all electrical equipment are designed for optimal performance on this supply. Harmonics are voltages and currents which have frequency components that are integers multiple of thefundamental frequency – pollutingthe pure sinusoidal waveform [1]. Power electronics such as those used in rectifiers, variable speed drives, UPS, lighting dimmer switches, televisions and hosts of other equipment, draw current in a non-sinusoidal fashion. This non-sine current interacts with the mains supply and distorts the voltage to a greater or lesser degree depending upon the strength or weakness (fault level) of the supply [2].

Nonlinear loads are not necessarily the only cause of harmonics within the supply system. Generators and the distribution system itself can produce harmonics. Some examples of nonlinear loads that generates harmonic currents are computers, fax machines, printers, refrigerators, TVs and electronic lighting ballasts. If a generator produces a non-ideal sinusoidal waveform, the voltage waveform can contain a certain amount of harmonics [3]. In the distribution system, generators are capable of producing harmonics due to magnetic core saturation. The generator used for the purpose of this analysis is a 3.5 kVA, at F (43.804 Hz), JINLING 3600 with manufacturer's data as: voltage 220V, frequency 50Hz, p.f 1 fuel type petrol.

#### II. BASICS OF FILTER DESIGN[4]

On the a-c side of the converters, harmonic filters are installed which limit the flow of harmonic in the a-c system, which otherwise may cause interference with telecommunication circuits and produce other deleterious effects.

#### III. CHOICE OF FILTER O[5]

In order to develop the fundamental ideas of filter design we will consider the L-C-R single tuned type. For the L-C-R elements in series, and with the arm at resonant frequency  $f_n$  (and where)  $\omega_n = 2\pi * f_n$  [6].

$$\omega_n^2 LC = 1 \tag{1}$$

by definition

$$Q = \frac{\omega_n L}{R}$$
 (2)

on substitution

$$R = \frac{1}{\omega_n CQ}$$
 (3)

Let  $\delta$  be per unit detuning expected from all causes, expressed as an effective frequency variation from the nominal [6].

$$Y_a = \frac{1}{Z_a} = G_a + jB_a \tag{4}$$

$$Y_f = \frac{1}{Z_f} = G_f + jB_f$$
 (5)

At an approximate near resonance,

$$G_f = \frac{1}{R(1 + 4Q^2 \delta^2)}$$
 (6)

$$B_f = \frac{-2Q\delta}{R(1+4Q^2\delta^2)} \tag{7}$$

$$Q = \frac{\omega_n L}{R}$$

$$\delta = \frac{\omega - \omega_n}{\omega_n}$$
(8)

 $\omega_n = 2\pi x$  nominal resonance frequency  $\omega = 2\pi x$  actual frequency

$$\therefore Y_T = (G_a + G_f) + j(B_a + B_f)$$
(9)

$$V_{n} = \frac{I_{n}}{Y_{T}}$$

$$\therefore |V_{n}| = \frac{I_{n}}{\left[\left\{G_{a} + \frac{1}{R(1 + 4Q^{2}\delta^{2})}\right\}^{2} + \left\{B_{a} - \frac{-2Q\delta}{R(1 + 4Q^{2}\delta^{2})}\right\}^{2}\right]^{1/2}}$$
(10)

Some interesting cases arise from equation 1.10 depending on the a-c system impedance.

#### IV. THE CHOICE OF CAPACITOR

The capacitor bank is the most costly item in the filter and usually accounts for about 60% of the filter cost, depending on the var rating of the capacitor bank. After making a decision on the amount of vars to be supplied by the filter circuits at fundamental frequency, the total capacitance of the filter bank is calculated as shown below [7].

$$MVAR = 2\pi * f * C * V^2 * 10^{-6}$$

$$C = \frac{MVAR * 10^6}{2\pi * f * V^2}$$

(11)

Having fixed the total value of capacitance it may be equally divided in all the branches. Usually the bulk of capacitance is not put as one unit but it is divided into smaller units and these are connected in series-parallel combinations to give the desired

capacitance. Below, the method to divide the capacitance in smaller units is given.

#### V. INDUCTOR RATING AND DESIGN [8]

Having fixed the value of the capacitor in the arm it is easy to calculate the value of reactor required to produce series resonance at that particular harmonic frequency. For resonance at the  $n^{th}$  harmonic

$$X_{L(1)} = \frac{X_{C(1)}}{n^2} \tag{12}$$

where.

 $X_{L(1)} = \text{fundamental reactance of inductor} \ X_{C(1)} = \text{fundamental reactance of capacitor}$ 

$$L_n = \frac{1}{\omega^2 * n^2 * C} \tag{13}$$

and the fundamental kVA rating of the reactor

$$kVA = \omega CE^2 \frac{n^2}{(n^2 - 1)} \tag{14}$$

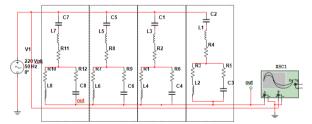
The main considerations in the design of an inductor are the impulse test voltage and the power loss. The most satisfactory arrangement is to have the impulse test voltage equal to that for the a-c line.

## VI. DEVELOPMENT OF DOUBLE TUNED LOW-PASS ACTIVE SHUNT FILTER CIRCUIT[9]

Assume that this reactive power is to be supplied by the filter circuits, connected to the primary side.

$$MVAR = 2\pi * E^2 * f * C * 10^{-6}$$

$$\therefore C = \frac{MVAR*10}{2*\pi*275*275*6.0}$$



**Fig.1:** 3.6 A developed section of double tuned low-pass active shunt filter circuit for simulation [9].

Single tuned arms are provided for the 5<sup>th</sup>, 7<sup>th</sup>, 11<sup>th</sup> and 13<sup>th</sup> harmonics. The fundamental frequency of a high-pass are nearly that due to its capacitor alone (with remaining components short circuited). But in the case of single-tuned arms, the voltage at the capacitor terminal is somewhat higher than the system voltage. The relationship for individual harmonics is

$$\frac{V_{C1}}{V_{n1}} = 1 + \frac{1}{n^2 - 1} \tag{15}$$

where.

 $V_{c_{\rm l}} = {\rm fundamental\ frequency\ voltage\ of}$  capacitor

 $V_{n_1}$  = fundamental frequency system voltage

n = harmonic number. This gives the following numerical values: when,

$$n = 1, 2, 3, 4, 5, 6, 7, 8, 9, 10, 11$$

The corresponding values of C, L and R will be computed thus:

$$\frac{V_{C1}}{V_{n1}}$$
 = For various values of *n*:

Therefore using these values, the capacitance of each arm can be reduced accordingly. We now turn our attention to the design of the inductive and resistance components. The following relationship applies for the value of reactance

$$X_{Ln} = \frac{1}{n^2} X_{nC}$$
 (16)

when.

 $X_{Ln}$ = reactance of reactor at fundamental frequency

 $X_{Cn}$  = capacitor reactance at the fundamental frequency

n = harmonic number.

$$2\pi f * L_n = \frac{1}{2\pi f_n * C}$$

$$L_n = \frac{1}{\omega_n^2 C} \qquad (17)$$

The quality factor of the coil is given by

$$Q = \frac{\omega_n L_n}{R_n}$$
 (18)

where

 $\omega_n = 2\pi * \text{normal resonance frequency.}$ 

The optimum value of Q as derived as in

$$Q = \frac{1 + Cos\phi_a}{2\delta + Sin\phi_a} \tag{19}$$

where,

 $\Phi_a$ = system impedance angle a  $\delta$  = total detuning expected due to all causes.

The presented example of the designed double-tuned filter demonstrates the generic algorithm usefulness as a tool for optimization for filter parameters without the need for unnecessary simplifications that may lead to erroneous solutions. The use of generic algorithm allows maximum filter currents/voltages across the filter components.

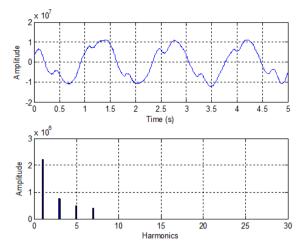
Efficient filter performance should give a distortion value of less than 5% at a-c system terminal. The effectiveness of the filters in eliminating a particular harmonic depends largely on the harmonic impedance of the a-c system at that frequency. But at present the harmonic impedance of a-c system cannot be computed or measured accurately and reliably at major harmonics. So usually it is assumed that the a-c system impedance can have any magnitude but a maximum impedance.

**Table (1)(a)** Power source 3.5 kVA, F (43.804 Hz), before loading [9]

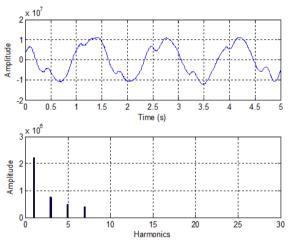
serore routing [7]										
				Magnitude	Phaseangle					
ham	time	(Voltage)	(Voltage)	(Voltage)	before filtering					
no	(sec)	a <sub>ca</sub> , (%)	b <sub>ss</sub> , (%)	C., %	(rad)	scope_data				
1	0	1895130.78	1141118.73	2212164.70	0.541985613	6800000				
2	0.01	0.00	0.00	0.00	0	6600000				
3	0.02	694397.69	297808.68	755564.80	0.40514685	6600000				
4	0.03	0.00	0.00	0.00	0	6600000				
5	0.04	446610.57	199096.51	488978.96	0.419351071	6600000				
6	0.05	0.00	0.00	0.00	0	6600000				
7	0.06	346852.19	154009.83	379506.88	0.417871103	6600000				
8	0.07	0.00	0.00	0.00	0	6800000				
9	0.08	284898.38	153378.68	323561.59	0.49386479	6600000				
10	0.09	0.00	0.00	0.00	0	6600000				
11	0.1	264992.47	130502.70	295384.43	0.457611126	6600000				
12	0.11	0.00	0.00	0.00	0	660000				
13	0.12	217863.88	126348.67	251850,47	0.525541249	6600000				
14	0.13	0.00	0.00	0.00	0	6600000				
15	0.14	187247.33	121925.44	223444.34	0.577180776	6600000				
16	0.15	0.00	0.00	0.00	0	6600000				
17	0.16	130519.73	112058.09	172024.46	0.709438559	6600000				
18	0.17	0.00	0.00	0.00	0	6600000				
19	0.18	107524.52	98205.78	145622.45	0.740133178	6400000				
20	0.19	0.00	0.00	0.00	0	6400000				
21	0.2	96619.19	87612.38	130426.98	0.736548553	6400000				
22	0.21	0.00	0.00	0.00	0	6400000				
23	0.22	91737.95	43560.77	101554.88	0.443317087	640000				
24	0.23	0.00	0.00	0.00	0	620000				
25	0.24	81743.99	38858.77	90510.13	0.443751528	6200000				

**Table (2)** Power source 3.5 kVA, F (43.804 Hz), before filtering

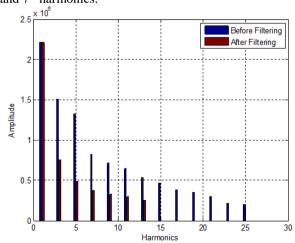
	before intering								
				Magnitude	Phaseangle				
ham	time	(Voltage)	(Voltage)	(Voltage)	after filtering				
no	(sec)	a <sub>ca</sub>	b <sub>sa</sub>	Caa, %	(rad)	scope_data			
1	0	2598777.76	1150992.83	2842257.97	0.41693202	680000			
2	0.01	0.00	0.00	0.00	0	660000			
3	0.02	828384.15	328301.65	891068.05	0.37732623	660000			
4	0.03	0.00	0.00	0.00	0	660000			
5	0.04	470467.69	264435.49	539690.63	0.51206229	660000			
6	0.05	0.00	0.00	0.00	0	660000			
7	0.06	337176.13	200629.98	392352.04	0.53675722	660000			
8	0.07	0.00	0.00	0.00	0	680000			
9	0.08	257663.57	152281.53	299299.49	0.53378236	660000			
10	0.09	0.00	0.00	0.00	0	660000			
11	0.1	241702.13	145644.95	282192.09	0.54231464	660000			
12	0.11	0.00	0.00	0.00	0	660000			
13	0.12	203652.78	113977.75	233378.20	0.51023482	660000			
14	0.13	0.00	0.00	0.00	0	660000			
15	0.14	156743.89	102925.76	187516.29	0.58103538	660000			
16	0.15	0.00	0.00	0.00	0	660000			
17	0.16	148662.93	61974.19	161063.55	0.39497052	660000			
18	0.17	0.00	0.00	0.00	0	660000			
19	0.18	122543.50	37964.44	128289.55	0.30042667	640000			
20	0.19	0.00	0.00	0.00	0	640000			
21	0.2	116742.87	30693.21	120710.28	0.25709461	640000			
22	0.21	0.00	0.00	0.00	0	640000			
23	0.22	103213.56	30526.67	107633.25	0.28756437	640000			
24	0.23	0.00	0.00	0.00	0	620000			
25	0.24	98394.73	25273.60	101588.77	0.25142399	620000			



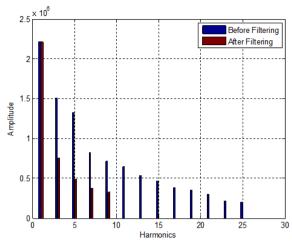
**Figure 2:** Power source 3.5 kVA, processed waveform and harmonic before loading with a non-linear load.



**Figure3:** Processed waveform and harmonic after filtering with a of power source 3.5kVA, waveform without load using the series active power filter harmonic spectrum showing the fundamental, 3<sup>rd</sup>, 5<sup>th</sup> and 7<sup>th</sup> harmonics.



**Figure4:** Harmonic spectrum of Power source 3.5kVA using series active power filter with passive filter shows the combined effect of the and after filtering showing  $1^{st} - 13^{th}$  harmonics



**Figure5:** Harmonic spectrum of Power source 3.5kVA using double tuned low-passactive shuntfilter shows the combined effect of the and after filtering showing  $1^{st} - 9^{th}$  harmonics

### VII. DISCUSSIONS: POWER SOURCE 3.5 KVA, F (43.804 HZ)

- 1. Figure 2 shows power source 3.5kVA waveform without load, the harmonic spectrum showing the fundamental frequency up to 7<sup>th</sup> harmonics, generators are expected to damped out all harmonics, but for the fact that the frequency (43.804 Hz at a rating of 3.5 kVA) at which the measurement was taken it was found out lower than the normal frequency (50 Hz) hence the associated harmonics found out when it was not loaded.
- Figure 3 shows the V<sub>THD</sub> results before / after simulation using the series shunt active filter 8.04%, 5.24% which is 0.24% higher than the expected IEEE standard of 5% after simulation. The spectrum shows 1<sup>st</sup> 7<sup>th</sup> harmonic.
- Figure 4 shows harmonic spectrum of Power source 3.5kVA using series active power filter with passive filter shows the combined effect of the and after filtering showing 1<sup>st</sup> 13<sup>th</sup> harmonics
- 4. Figure 5 shows the  $V_{THD}$  results before / after simulation using the double tuned low- pass active shunt filter 6.24%, 1.51% which is 3.49% less than the expected IEEE standard of 5% after simulation. The spectrum shows  $1^{st} 9^{th}$  harmonic.
- 5. The frequency of the power source 3.5kVAis 43.804Hz which less than the actual standard 50Hz of operation, this could be a possible reason for the high harmonic present in this power source.

#### VIII. CONCLUSION

An efficient but simple technique has been developed for improvement of power quality disturbances during faulty conditions using double

tuned low-pass active shunt filter circuit for simulation of harmonic in the electrical power system. The proposed model of the system is simulated in the MATLAB/Simulink environment. The results show that double tuned harmonic filter is effective in reducing the magnitude and frequency of voltage spikes significantly.

The presence of a parallel  $C_3 - L_3$ ,  $C_4 - L_4$ ,  $C_6 - L_6$  and  $C_8 - L_8$  circuit will result in circulating harmonic currents, which in the case of the capacitor current can exceed the fundamental current. The magnitude of such circulating currents can be controlled by lowering the Q value that is increasing the resistance or installing a resistor R in the circuit. The effectiveness of the filters in eliminating a particular harmonic depends largely on the harmonic impedance of the nonlinear load used in the system at that frequency.

It may be concluded that more economic and efficient designs for the filters may be obtained by better knowledge of and greater experience in the following:

- 1. A-C system parameters, especially the network impedance at harmonic frequencies.
- 2. Reactive power capabilities of the a-c system.
- 3. Correlation between the calculated distortion values and those experienced in practice, and the determination of acceptable levels.
- 4. Generators and the distribution system itself can produce harmonics.

#### **REFERENCES:**

- [1]. Dugan R, McGrannaghan M, Santoso S, and Beatty H, 2005, "A conceptual power quality monitoring technique based on multi-agent systems". In: 37th Power Symposium, North American, pp.358–363.
- [2]. Blaabjerg F. Sankaran, C. and Meier A., 2006, 'Understanding Power Quality Problems: Voltage Control and Interruptions'. New York: http://pt.wikipedia.org/wiki/Transformada inversa delaplace
- [3]. Vaez-Zadeh, E. C Shireen W and Pitel I, IEEE-519, 2002, "IEEE Recommended Practices and Requirements for Harmonic Control in Electric Power Systems", IEEE Transactions on Power Delivery, vol. 8, no. 3, pp.719-72
- [4]. Lawler J. S, Nagpal M., McGranaghan, M. F., Grebe, T. E. and Samotyj, M. 2004 "Design and Analysis of Harmonic Filters Using MATLAB" by Ahmed Hassan & Azeem Talib U.E.T Lahore, http://followscience.com/content/212834/electricalpower-systems-quality-second-edition

- [5]. Akagi H, 2013, "Control Strategy and Site Selection of a Shunt Active Filter for Damping of Harmonic propagation in Power Distribution Systems" IEEE Transactions on Power Delivery, Vol. 12, No 1.
- [6]. Rhyne G.W., Steer M.B., and Bates B.D., 2008, "Frequency-Domain Nonlinear Circuit Analysis Using Generalized Power Series," IEEE Transactions on Microwave Theory and Techniques, Vol. 36, No. 2, pp. 379-387
- [7]. Adeyemo D.O. and Adeleke S.A, (2005), "System and Method for Determining Power Quality Contributions from Non-linear Loads," in Proceedings of the IEEE Industry Applications Society Annual Meeting (IAS 2005), Kowloon, Hong Kong, pp. 2456-2463
- [8]. Sumit Kanwar, Om Prakash Mahela and Devendra Mittal, 2013, 'Power Quality Improvement in Faulty Conditions using Tuned Harmonic Filters', IOSR Journal of Electrical and Electronics Engineering, e-ISSN: 2278-1676,p-ISSN: 2320-3331, Volume 6, Issue 5, www.iosrjournals.org
- [9]. Eyenubo O. J, 2015, "Power Quality Analysis: The Reduction of Harmonic Contents in Power Systems USING Electronic Filter',Ph.D Thesis, University of Benin Department of Electrical and Electronic Engineering, pp. 101 108

<sup>\*</sup> Dr. Eyenubo Ogheneakpobo Jonathan. "Power Quality Analysis of Harmonics in 3.5kVA Single Phase Generator using Electronic Filter in Matlab/Simulink Environment." International Journal Of Engineering Research And Applications (IJERA), vol. 08, no. 01, 2018, pp. 45–49.