

Robust Nonlinear Channel Equalization using WNN trained by Symbiotic Organism Search Algorithm

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ABSTRACT

In the present world of 'Big Data,' the communication channels are always remaining busy and overloaded to transfer quintillion bytes of information. To design an effective equalizer to prevent the inter-symbol interference in such scenario is a challenging task. In this paper, we develop equalizers based on a nonlinear neural structure (wavelet neural network (WNN)) and train it's weighted by a recently developed meta-heuristic (symbiotic organisms search algorithm). The performance of the proposed equalizer is compared with WNN trained by cat swarm optimization (CSO) and clonal selection algorithm (CLONAL), particle swarm optimization (PSO) and least means square algorithm (LMS). The performance is also compared with other equalizers with structure

based on functional link artificial neural network (Trigonometric FLANN), radial basis function network (RBF) and finite impulse response filter (FIR). The superior performance is demonstrated on equalization of two non-linear three taps channels and a linear twenty-three taps telephonic channel. It is observed that the performance of the gradient algorithm based equalizers fails in the presence of burst error. The robustness in the performance of the proposed equalizers to handle the burst error conditions are also demonstrated.

Keywords: Wavelet Neural Network, Symbiotic Organism Search Algorithm, Channel Equalization, Burst Error.

INTRODUCTION

With the advancement in computing and signal processing techniques in this era of big data, everyone

wishes to share huge volume of information (movies, audio-video files, graphic games etc.) to many users simultaneously. This increases the load on the bandwidth limited communication channels causing intersymbol interference (ISI) [1]. In order to neutralize the effect

of ISI, a signal processing module is introduced at the receiver known as channel equalizer [2, 3]. Since in recent past the volume of information sharing has increased, the interference level in the channel has simultaneously increased to many folds. Thus the problems that equalizers have to take up with these crowded channels are getting more complicated than ever. Adaptive channel equalizer was first proposed by Lucky in 1965 [4]. In the original form of the equalizer, there received signal passed through a tapped delay line followed by an adaptive linear combiner. The adjustable weights of the model are trained by least mean square (LMS) [5], Normalized LMS [6], recursive least square (RLS) [7], fast RLS [8], square-root RLS [9] etc. However these linear models with gradi-

ent descent algorithms most of the time fails to equalize the channels with deep spectral nulls [10]. Therefore neural network based equalizers are introduced which formulates nonlinear decision boundaries to handle complex classification tasks [11]. The first of this kind was Multilayer Perceptron (MLP) trained by Back Propagation (BP) algorithm [12, 13]. In order to handle equalization of QAM signal complex domain MLP and extended BP is proposed [14]. To improve the convergence speed of MLP, natural gradient (NG) descent algorithms based equalizers are proposed [15], which provide faster convergence than BP.

In order to reduce the complexity of multilayer structure, later on, single-layer structures like Radial Basis Function Neural Network (RBF) [16], Functional Link Artificial Neural Network (FLANN) [17, 18] are employed for equalization. For QAM equalization the complex valued RBF are developed [19, 20]. Based on the type of nonlinearities associated with the channel the Chebyshev artificial neural networks [21, 22], Legendre neural networks [23, 24] are introduced which become popular in the last decade. Recently the hybrid FLANN

(combination of FIR structure with FLANN) [25] and pipelined FLANN are proposed by [26] Zhao et al. which provide accurate results and efficient hardware architecture than the traditional FLANN.

In order to handle the time-varying phenomenon of fading channels (a dynamic system with its coefficients varying with time) Kechriotis et al. proposed the use of Recurrent Neural Network (RNN) [27]. The RNN structure comprises an IIR filter with feedback system, along with nonlinearities associated with the neurons [28]. The RNN outperforms the MLP and RBF structures [29], and can be used for both trained as well as blind equalization [30]. Choi et al. in 2005 used Kalman Filter to train the weights of RNN equalizer [31]. Recently Wang and Huang analyzed the convergence of RNN trained by extended Kalman Filter [32] considering the covariance of measurement noise (R) and process noise (Q) into account.

Patra and Mulgrew used fuzzy logic for adaptive nonlinear equalization [33]. The extended Kalman filter is used by Li and Er to train the fuzzy neural equalizer [34]. Recently a neuro-fuzzy system based on clustering and gradient techniques is used for equalization [35]. The review article by Burse et al. highlights the detail of the evolution of various neural network models for channel equalization [10].

Wavelet Neural Network (WNN) [36] consists of a feed-forward single hidden layer neural network with a linear combiner at the output. The activation function used at the hidden layer neurons are mother wavelets. The wavelet function with dilation and translation parameters can effectively approximate nonlinear functions or formulate nonlinear decision boundaries for

classification task [37]. In the last decade the WNN has been widely applied for nonlinear system identification [38], prediction of Chaotic time series [39], rainfall prediction [40], wind power forecasting [41] cancer classification of microarray gene expression data [42], human lower extremity joint moment prediction [43]. Pradhan et al. [44] applied the WNN for communication channel equalization. They reported the superior performance of WNN trained by extended Kalman filter (EKF) over multi-layered perceptron (MLP) and radial basis function neural network (RBFNN).

The recursive derivative based algorithms like LMS, RLS, BP, NG, EKF discussed above follows the hill-climbing approach with fixed step size to achieve optimality. During the process of moving towards optimality if they come across the local optimums they get stuck to it and thus the process leads to inaccurate modeling. Nature-inspired algorithms [45] use a group of elements (known as population) who combined use their intelligence to explore the search space and determine the optimality. Therefore these algorithms ensure greater probability to achieve accurate modeling and equalization.

In the last decade number of nature inspired algorithms have been used for channel equalization including : genetic algorithm [46, 47], simulated annealing [48], bacterial foraging optimization [49, 50], differential evolution [51], artificial immune systems [52, 53], artificial bee colony [54], particle swarm optimization [55, 56, 57], shuffled frog leaping algo-

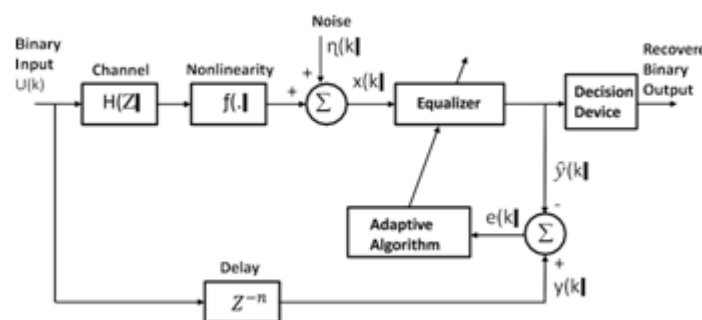


Figure 1: Block diagram of an adaptive communication channel equalizer.

algorithms for channel equalization. The performance of the proposed equalizer is compared with other WNN based equalizers where training is carried out by CSO, CLONAL, PSO and LMS. The performance is also compared with Trigonometric FLANN, RBF and FIR based equalizers trained with above mentioned nature inspired and gradient based algorithms. Extensive simulation studies of the above equalizers are demonstrated on two non-linear and a linear twenty-three taps telephonic channel. Many times the gradient algorithm based equalizers fail to accurately model the channel characteristics in the presence of burst error. The robustness of the proposed equalizers to handle burst error conditions are also analyzed.

1. Adaptive Communication Channel Equalization

The block diagram of an adaptive communication channel equalizer is shown in Figure 1. The input signal $U(k)$ is random binary in nature, taking values in the form either -1 or +1. The input pass through a nonlinear communication channel having transfer function $H(z)$ and associated non-linearity $f(.)$. While passing through the channel the signal gets contaminated with additive white Gaussian noise (AWGN) $\eta(k)$. The output obtained at the receiver is

algorithm [58], krill herd algorithm [59]. Recently published survey article by Gotmare et al. [60] highlight the use of swarm and evolutionary techniques for system identification, filter de-

$$x(k) = f \left[\sum_{j=1}^N h(j)u(k-j) \right] + \eta(k) \quad (1)$$

sign and channel equalization. Cheng and Prayogo developed a new meta-heuristic optimization algorithm Symbiotic Organisms Search (SOS) [61] by observing the interaction strategies adopted by various organisms to survive and propagate in the ecosystem. The benefit of SOS is, it's parameter-free. Only common control parameters that are population size and a maximum number of function evaluations are to be adjusted. Therefore the performance of SOS is consistent for different optimization problems. The authors have demonstrated the superior performance of SOS on unconstrained optimization of twenty-six mathematical functions and four optimal design problems of structural engineering. Based on the recent trend of research in this paper we develop equalizers based on wavelet neural network

trained by symbiotic organisms search algorithm. We have not come across any literature on the training of WNN structure with nature inspired

where $h(i)$, $i = 1, \dots, N$ represent the impulse response of the channel, $u(k)$ and $\eta(k)$ are the input data sample and noise at k th instant, with $k = 1, \dots, K$. The K is the total number of input sample applied.

At the receiver, the equalizer has a transfer function $W(z)$, where $W(z) = 1/H(z)$ (zero force equalization to compensate inter-symbol interference (ISI) of the channel). In nonlinear channel, the equalizer has to neutralize the ISI as well as the nonlinearity associated with the channel. The output of the equalizer at k th instant is given

$$\hat{y}(k) = \sum_{j=1}^Q x(k-j)w(j) \quad (2)$$

where Q is the order of the equalizer and $w(j)$ is the associated adaptive weight vector to it. The desired output $y(k)$ is obtained

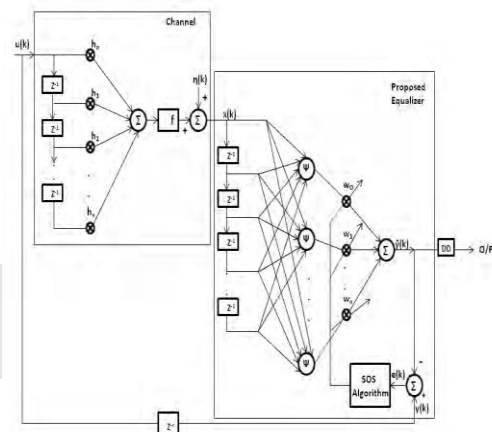


Figure 2: Block diagram of proposed WNN trained by SOS equalizer.

at the receiver by delaying the input sample $u(k)$ by n samples. In practice the delay $n = Q/2$. The error signal $e(k)$ generated at the receiver is given by

$$e(k) = y(k) - \hat{y}(k) \quad (3)$$

The objective is to recursively reduce the errors so that $\hat{y}(k)$ approaches $y(k)$. The decision device recovers the desired signal which is also used for evaluating the bit error rate (BER) performance. The decision device is represented by a symmetric hard limit transfer function and its output $d(k)$ given by

$$d(k) = \begin{cases} 1 & \text{for } y(k) \geq 0 \\ -1 & \text{otherwise} \end{cases} \quad (4)$$

2. Proposed WNN trained by SOSEqualizer

Wavelet Neural Network(WNN)

The WNN [36] is a single hidden layer neural network with a weighted linear combiner at the output. The nonlinear activation function used at hidden layer neuron nodes are mother wavelets. Pradhan et al. [44] initially applied the WNN structure for communication channel equalization. They reported

the superior performance of WNN trained by extended Kalman filter (EKF) equalizer over multi-layered perceptron (MLP) and radial basis function neural network (RBFNN) based equalizers. The estimated output of the WNN equalizer is given by

$$\hat{y} = \sum_{j=1}^k W_j \Psi_j(x)$$

where W_j is the weight between j^{th} hidden layer and output layer. k is the number of hidden units. The $\Psi_j(x)$ is multi-dimensional wavelet associated with j^{th} unit of hidden layer. It is constructed with the product of m scalar wavelets with x .

The block diagram of proposed Wavelet neural network (WNN) trained by Symbiotic organism search (SOS) equalizer

$$\Psi_j(x) = \prod_{i=1}^m \psi\left(\frac{x - b_i}{c_i}\right)$$

is shown in Figure.1. The various component of the equalizer is outlined below. where ψ is the mother wavelet and b_i and c_i are translation and dilation parameters of the mother wavelet function. The x is input vector i.e. the output of the channel and m is number of inputs to the WNN structure.

In this case we have selected the same 'Mexican Hat' as mother wavelet function which has already been reported by Pradhan et al. [44] to provide better results for equalization purpose. The 'Mexican Hat' based wavelet function is given by

$$\psi\left(\frac{x-b_i}{c_i}\right) = \exp\left(-\frac{x-b_i}{c_i}\right) \exp\left(-\frac{x-b_i}{c_i}\right) \exp\left(-\frac{x-b_i}{c_i}\right) \quad (7)$$

The estimated WNN equalizer output $\hat{y}(k)$ is compared with the desired target during training to produce error

$$e(k) = y(k) - \hat{y}(k) \quad (8)$$

a significant amount of benefit while interacting with j (grass). But when j (grass) interacts with i (horse) it may get only nominal or adequate benefit. Therefore the scaling term 'benefit factor' is used which randomly takes the value as either 1 or 2 (corresponds to partial or full benefit by the organism).

The MV is termed as 'Mutual Vector' which describes the similarity of characteristic or average characteristics of organism i and j . The part of equation $(O_{\text{best}} - MV * BF_2)$ reflects

The estimated WNN equalizer output $\hat{y}(k)$ is compared with the desired target during training to produce error

The aim is to train the weights of the WNN equalizer recursively so that the $\hat{y}(k)$ approaches $y(k)$ which is essential for the equalization. In this paper, the Mean Square Error (MSE) is minimized on every iteration with symbiotic organism search (SOS) algorithm.

Symbiotic organism search (SOS) algorithm

The Symbiotic organism search (SOS) is a nature inspired metaheuristic algorithm recently developed by Cheng and Prayogo [61]. It is based on the interaction strategies adopted by various organisms to survive and propagate in the ecosystem. The algorithm avoids the fall to the local minima and thereby avoids the premature convergence. The flowchart for the algorithm is shown in Figure.3. The natural organisms in an ecosystem normally try to live together. The process of living together is termed as 'Symbiosis'. It's essential for the organisms to stay together as the organisms of different species dependent on each other either for their survival or for common growth. The fundamental symbiotic relationships observed in the natural organisms are mutualism, commensalism, and parasitism. These three natural phenomena are encoded in the form of SOS algorithm. The in-detail steps are outlined below :

Mutualism phase

In this phase, the interaction between two organisms of different species results in benefiting both the organisms. Therefore it is also termed as 'Phase of benefits'. As shown in Figure.4, the

relationship between horse and grass is mutualism. By grazing the grass horse fill it's stomach, also it en-reaches the growth of grass.

In SOS algorithm consider an ecosystem \mathbf{O}

$= [O_1, O_2, \dots, O_M]$. The mutualism between two organisms $[O_i, O_j]$ \mathbf{O} and $i \in \{1, 2, \dots, M\}$

(i.e. consider O_i as horse and O_j as grass) is given by

$$O_{i-new} = O_i + \text{rand}(0,1) * (O_{best} - MV * BF_1) \quad (9)$$

$$O_{j-new} = O_j + \text{rand}(0,1) * (O_{best} - MV * BF_2) \quad (10)$$

$$MV = \frac{O_i + O_j}{2} \quad (11)$$

where $\text{rand}(0,1)$ represents a random number in the range $[0,1]$. BF_1 , BF_2 , are termed as benefit factors which reflect the interaction between the organisms. Organism i (horse) may achieve towards the best organism. The organism O_{best} represents the organism which has the highest degree of adaptation to the ecosystem. Therefore, the principle obeys Darwin's theory of 'survival of the fittest'.

Finally, the organisms are updated with the above procedure only when their present evaluated fitness are better than the previous one.

Commensalism phase

In commensalism phase the interaction between two organisms benefit one, the other one neither gets benefit nor suffers from the relationship. In Figure.4, the interaction between Remora fish and Shark is commensalism. The tiny Remora attach itself behind the body of Shark and takes its leftover food, thus get benefited from the relation. The Shark do not get affected by the activity of the small fish. Also, the benefit received by Shark from the relationship is insignificant.

Consider two organisms $[O_i, O_j]$ \mathbf{O} and $i \in \{1, 2, \dots, M\}$. Consider O_i as Remora fish and O_j as Shark. The commensalism interaction between them is given by

$$O_{i-new} = O_i + \text{rand}(-1,1) * (O_{best} - O_j) \quad (12)$$

where $\text{rand}(-1,1)$ represents a random number in the range $[-1, 1]$. The term $(O_{best} - O_j)$ represents the beneficial advantage provided by O_j (Shark) to organism O_i (Remora fish) by increasing its survival advantage in the ecosystem.

Parasitism phase

In parasitism phase, the interaction between two organisms eliminates a weaker organism from the ecosystem. In Figure.4, the interaction between a mosquito and another bird is a phenomenon of parasitism. Once the mosquito bites the bird it creates parasite thrives in the bird's body. The germs inside the bird's body reproduce themselves and the bird may get affected by a

disease and possibly die. But if the bird is a healthier one and have good immunity power then it can cure itself and the parasite will get eliminated from that ecosystem.

Consider two organisms $[O_i, O_j]$ \mathbf{O} and $i \in \{1, 2, \dots, M\}$. The parasitism interaction between them is implemented as follows :

1. O_i is given the role of mosquito. It create a 'Parasite Vector' by cloning (duplicating) the organism O_i and randomly replacing selected dimension values with random numbers.
2. O_j is given the role of host bird.
3. Evaluate the fitness $f(\cdot)$ of 'Parasite Vector' and O_j .
4. If $f(\text{Parasite Vector}) < f(O_j)$, then O_j is replaced by the 'Parasite Vector'.



Figure 4: Diagram representing the process of symbiosis in an ecosystem.

put combiner are optimized using SOS algorithm. Therefore an ecosystem is initialized consist of the M number of organisms. Each organism has N dimension same as the number of a number of weights present in the WNN structure. Let the ecosystem be represented as

4. Performance Evaluation of WNN-SOSEqualizer

Channel Characteristics for Simulation

In order to evaluate the performance of the proposed WNN-SOS equalizer simulation studies are carried out on two benchmark channels. The channel 1 is taken from Pradhan et al.[44] and channel 2 is taken from Nanda et. a al. [52]. The transfer function of these channels are given by

$$\begin{aligned} \text{Ch1: } H_1(z) &= 0.2602 + 0.9298z^{-1} + 0.2602z^{-2} \\ \text{Ch2: } H_2(z) &= 0.3040 + 0.9030z^{-1} + 0.3040z^{-2} \end{aligned} \quad (15)$$

Each of the above mentioned channels are associated with two different kind of nonlinearities represented by

$$\begin{aligned} \text{NL1: } f_1(k) &= \tanh(g(k)) \\ \text{NL2: } f_2(k) &= g(k) + 0.2g^2(k) - 0.1g^3(k) \end{aligned} \quad (16)$$

Calculate desired output of the equalizer :

The desired output $y(k)$ of the equalizer is obtained by delaying the input signal $U(K)$ by 'n' samples.

5. Evaluate fitness and determine O_{best} : The output of the equalizer $\hat{y}(k)$ is calculated using (5) for every k th input sample and m th weight vector O_m . The obtained output is compared with the desired output $y(k)$ to produce error signal $e(k)$ using (8).

For m th organism (i.e. weight vector) the mean square error (MSE) is evaluated and used as its fitness function given by

K obtained by applying the input samples 4 unit delay. The training of the WNN structure weights with SOS algorithm is carried out for 200 iterations as per the procedure defined in Section

3.3. The obtained MSE during training is reported. Once the training is completed the testing of the equalizer is carried out using 1,00,000 input samples. The obtained BER is reported.

Comparative Models and Parameter Settings
The performance of the proposed equalizer is compared with other WNN equalizers trained by cat swarm optimization (CSO), CLONAL selection algorithm of artificial immune systems and particle swarm optimization (PSO). The parameter MSE_{om} settings of the heuristic algorithms for comparative analysis in

$k=1$
The objective is to minimize the MSE. Determine the O_{best} which has minimum MSE.

6. Perform Mutualism phase : Using the obtained O_{best} and equations (9)-(11) perform the mutualism phase as described in the Section 3.2.1.

7. Perform Commensalism phase : Similarly with the obtained O_{best} and equation (12) perform the commensalism phase as narrated in the Section 3.2.2.

8. Perform Parasitism phase:

Perform the parasitism phase as per the procedure described in the Section 3.2.3.

9. Stopping criteria : The organism (weight vector) which provide the minimum MSE corresponds to the best survival in the ecosystem. The steps 6-9 are repeated until a desirable minimum MSE is obtained.

Table 1: Parameter Settings for proposed WNN-SOS, WNN-CSO and WNN-CLONAL equalizer.

Para	WNN SOS	WNN CSO	WNN CLONAL	WNN PSO
Pop	20	20	20	20
Iter	200	200	200	200
Off-springs	Mutualism Commensalism Parasitism	Tracing Seeking	Cloning Hyper-mutation Selection	
Other	MR=0.9	Prob.	CI=100%	C1=2
Para-meters		SRD=20% c=2 $\zeta \in [0,1]$ $\eta \in [0.9,0.4]$	Prob.Hy=10%	C2=2 $r1 \in [0,1]$ $r2 \in [0,1]$

the identical environment are described in Table.1. The weight update procedure using CSO is narrated in [39] and that with CLONAL is outlined in [52]. The other comparative used are radial basis function (RBF) with 12 neurons in a single layer, the trigonometric FLANN where each term is expanded into two using $\sin(\pi x)$ and $\cos(\pi x)$ thereby producing 16 terms. The finite impulse response (FIR) filter structure is designed with eight taps.

Simulation Environment
The simulation studies of proposed WNN-SOS, comparative WNN-CSO and WNN-CLONAL equalizers are carried out in a MATLAB version R2011 platform on an Intel i7-3540M 3GHz CPU, with a 8GB RAM in Windows 8.1 (64-bit) environment.

RESULT AND DISCUSSIONS

The MSE curves obtained during training of the equalizers for channel 1 with nonlinearity 1 and nonlinearity 2 are shown in Figure.5(a) and Figure.5(c) respectively. Similarly, the MSE for channel 2 with nonlinearity 1 and nonlinearity 2 are shown in Figure.6(a) and Figure.6(c). In all the four case studies it is observed that the proposed WNN-SOS based equalizer provides faster convergence and lower MSE than that achieved by the WNN-CSO and WNN-CLONAL. Among the WNN-CSO and WNN-CLONAL the WNN-CLONAL equalizer has faster convergence but

WNN-CSO provides lower MSE.

The BER plots obtained during testing of the equalizers for channel 1 with nonlinearity 1 and nonlinearity 2 are shown in Figure.5(b) and Figure.5(d) respectively. For nonlinearity 1 the performance of proposed WNN-SOS is better than WNN-CSO which is better than that achieved by WNN-CLONAL. For nonlinearity 2 the performance of WNN-CLONAL and WNN-CSO are better up to 8dB after which the WNN-SOS provides better performance. In this case, the performance of WNN-CLONAL is marginally better than WNN-CSO. The BER plot of channel 2 with nonlinearity 1 and nonlinearity 2 are shown in Figure.6(b) and Figure.6(d) respectively. In both nonlinearities, the performance of the proposed WNN-SOS is better after 6dB. Prior to 6dB, the performance of WNN-CSO and WNN-CLONAL are better. However, the performance of the WNN-CLONAL for this channel becomes inferior to that achieved by SOS and CSO equalizers.

The performance of the proposed WNN-SOS equalizer with other comparative models for nonlinear channel 1 and 2 are demonstrated in Table 2. During training the settled value of MSE and Normalized MSE in dB are reported. During testing the BER value obtained at 20dB SNR are reported. It is observed that in SOS based training provides lower MSE is case of most of the structures (which are highlighted in bold letters). The WNN-SOS combination has the superior performance other other equalizers in both the nonlinear channels. It is also observed that for Channel 1 the performance of most of the equalizers is better for nonlinearity 2 compared to nonlinearity 1. But in case of Channel 2 the performance of the equalizers with Non-linearity 1 is better. For channel 1 the performance of WNN is followed by RBF, FLANN and FIR. For

Channel 2 the WNN is followed by FLANN, RBF and FIR. Among the training algorithms the SOS has greater accuracy followed by CSO then CLONAL/PSO and LMS.

5. WNN-SOS equalizer for Telephone Channel with 16-QAM

The example taken here is a telephone channel with 21 taps listed in Table.3. Earlier Chen et al. [62] designed multi-stage blind clustering equalizer for this channel. Here a pseudo-random binary sequence is generated of period 256 bits. The generated signal is contaminated by Gaussian noise of SNR 30 dB. Every four bit are converted into 16-QAM symbol. The equalizer with FIR filter comprised of 23 taps. Similarly, for proposed WNN-SOS equalizer number of neurons taken are 25. In RBF is designed with 25 neurons and

FLANN has 42 expansions (21 tap with each one expanded one sin and one cos term).

The constellation diagram of the noisy telephone channel output without equalization is shown in Figure.7 (a). Followed by the constellation output for the proposed WNN-SOS equalizer along with WNN-CSO and WNN-CLONAL are shown in Figure.7(b-d). The comparative performance of the proposed equalizer along with the other models for this telephonic channel is shown in Table.4. It is observed that the performance of the WNN based equalizers are superior followed by RBF, FLANN and FIR. Among the training algorithms the SOS provides lower MSE followed by CLONAL/PSO and LMS.

6. Ability of WNN-SOS Equalizer to Handle Burst Error

The recent US Patent [63] by Taher and Al-Banna focus on the usage of an equalizer to handle burst error. The burst error refers to the appearance of 'consistent ones' or 'consistent zeros' for certain duration of time. It happens in the physical channels as they are imperfect and cause an error at the receiver. The burst error in an equalizer may occur after the convergence of the algorithm and it always degrades the performance of the equalizer.

The simulation of a traditional FIR filter based equalizer trained by least mean square (LMS) algorithm response during the burst error is shown in Figure.9. The simulation is carried out on channel 1 with nonlinearity 1 in presence of AWGN 30 dB. After 100 samples of the input signal, the LMS is about to settle down when the burst error occurred for 20 samples (in the simulation from 10th bit to 120th bit all samples are taken as ones). During this period the error in the channel immediately increased to many folds and the algorithm again took a noticeable amount of time for convergence as shown in Figure.9(a). The NMSE representation of the same shown in Figure.9(b) which reveals the algorithm initially trying to achieve a convergence near about 30 dB around the 100th sample where the burst error occurred and the algorithm took the time to achieve the convergence again.

The block diagram shown in Figure.8 represents the implementation of the proposed WNN-SOS equalizer to handle the

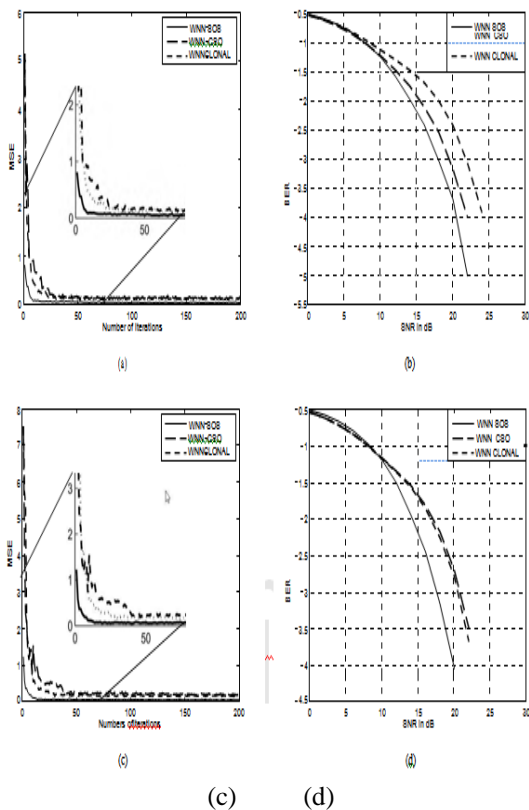


Figure 5: Simulation study of Channel 1 using proposed WNN-SOS with other comparative models : (a) MSE convergence during training with non-linearity 1 (b) BER obtained during testing with non-linearity 1 (c) MSE convergence during training with non-linearity 2 (d) BER obtained during testing with non-linearity 2.

burst error. The training is carried out with the evolutionary algorithms for 200 iterations. The same environment as mentioned above (i.e. burst error occurrence after 100th samples for duration 20 samples) is used for simulation. The Wilcoxon norm [64] is used to handle the burst error. The detailed mathematical formulation of the fitness function is given in [65].

The MSE curves obtained during training for channel 1 with nonlinearity 1 and nonlinearity 2 are shown in Figure 10 (a) and Figure 10(c) respectively. Similarly, the MSE for channel 2 with nonlinearity 1 and nonlinearity 2 are shown in Figure 11(a) and Figure 11(c). In all the four case studies it is observed that with the use of evolutionary learning the ability of the equalizer to handle the burst error has become much effective. It can effectively overcome the hurdles faced by LMS (as the population use collaborative epoch based learning of the weights). However compared to the MSE curves of normal condition (i.e. Figure 5(a)-5(c) for Ch1) in burst error condition

(Figure 10(a) and Figure 10(c)) the initial errors occurred in the training process are high (i.e. in Figure 5(a) the start level is 5 whereas in Figure 10(a) it's 8, similarly in Figure 5(c) the start level is 8 whereas in Figure 10(c) it's 18). The algorithm achieves the convergence to reduce the error. Almost similar observation is also found for Channel 2.

After the training is completed the testing is carried out for burst error conditions. Among the 1,00,000 testing samples, 60 samples were used as burst error conditions (20 samples after 10,000, then 20 samples after 40,000, then 20 samples after 60,000 were made ones). The BER plots obtained during this testing for channel 1 with nonlinearity 1 and nonlinearity 2 are shown in Figure 10(b) and Figure 10(d) respectively. Similarly, the BER for channel 2 with nonlinearity 1 and nonlinearity 2 are reported in Figure 11(b) and Figure 11(d) respectively. Compared to the normal conditions of BER (Figure 5(b), Figure 5(d), Figure 6(b), Figure 6(d)) in burst error condition the performance of WNNCSO and WNNCLONAL equalizers have been affected (Figure 6(b), Figure 6(d), Figure 11(b), Fig-

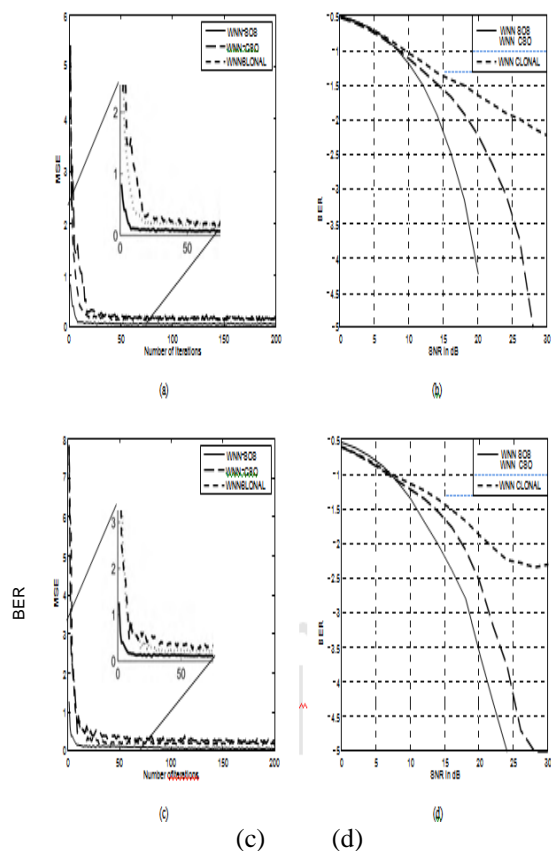


Figure 6: Simulation study of Channel 2 using proposed WNN-SOS with other comparative models : (a) MSE convergence during training with

non-linearity 1 (b) BER obtained during testing with non-linearity 1 (c) MSE convergence during training with non-linearity 2 (d) BER obtained during testing with non-linearity 2.

ure.11(d)) whereas the WNN-SOS equalizer performance still remains better. Therefore the WNN-SOS equalizer is a potential candidate under burst error scenario.

CONCLUSION

In this paper, a new channel equalizer is proposed by training the weights of a wavelet neural network structure with recently developed symbiotic organisms search algorithm. The performance of the proposed equalizer is compared with Trigonometric FLANN, RBF and FIR based structures trained by CSO, CLONAL, PSO and LMS algorithm. The performance of the equalizers is accessed over two nonlinear channels for regular noisy conditions and burst error scenarios. The performance is also demonstrated on a twenty-one taps telephonic channel. The MSE and BER results in all simulated scenarios reveal that the WNN-SOS equalizer provides superior performance than that achieved by the other equalizers. Therefore, it is up to you to try this hybrid combination of WNN-SOS to solve complex identification and nonlinear classification problems.

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Table 2: Comparative performance analysis of the proposed WNN-SOS equalizer with other neural network and FIR based equalizers for both the non-linear channels.

Structure	Algorithm	Training		Testing	Training		Testing
		MSE	NMSE (dB)	BERvalue 20 dBSNR	MSE	NMSE (dB)	BERvalue 20 dBSNR
Channel 1		Non-linearity 1			Non-linearity 2		
WNN	SOS	0.1021	-28.6240	-3.7100	0.1560	-28.8321	-4.1232
	CSO	0.2114	-25.4210	-3.2332	0.2525	-24.3210	-2.8530
	CLONAL	0.2515	-23.3762	-2.5604	0.2620	-23.8020	-2.6126
	PSO	0.2604	-23.124	-2.5412	0.2633	-23.2126	-2.6324
	LMS	0.3510	-21.8022	-2.3020	0.3622	-21.8085	-2.4212
FLANN	SOS	0.1430	-28.3102	-3.5212	0.2120	-28.3218	-3.8236
	CSO	0.2245	-24.6810	-3.1024	0.2411	-24.7064	-2.9166
	CLONAL	0.2412	-23.3546	-2.4105	0.2513	-23.3712	-2.4212
	PSO	0.2652	-22.9020	-2.3320	0.2688	-23.2012	-2.3410
	LMS	0.3632	-21.2255	-2.0243	0.3650	-21.0115	-2.1210
RBF	SOS	0.1356	-28.3530	-3.6140	0.2094	-28.4125	-4.0135
	CSO	0.2215	-25.2104	-3.1614	0.2366	-25.6124	-2.8842
	CLONAL	0.2414	-23.3633	-2.5245	0.2518	-23.4012	-2.5446
	PSO	0.2632	-22.9255	-2.4432	0.2673	-23.1014	-2.5020
	LMS	0.3576	-21.2570	-2.1340	0.3642	-21.754	-2.3110
FIR	SOS	0.3240	-23.5014	-2.8246	0.3267	-23.7125	-2.9137
	CSO	0.3634	-21.0350	-2.5411	0.3735	-22.0024	-2.6325
	CLONAL	0.3685	-20.2013	-2.4628	0.3822	-21.4320	-2.6405
	PSO	0.3710	-20.1105	-2.2014	0.3884	-20.1836	-2.1225
	LMS	0.4012	-19.8620	-1.9940	0.4168	-19.9448	-2.0010
Channel 2		Non-linearity 1			Non-linearity 2		
WNN	SOS	0.0912	-30.0114	-4.2032	0.1114	-28.1056	-3.6010
	CSO	0.1322	-28.3206	-2.5216	0.1426	-27.1136	-2.5024
	CLONAL	0.1284	-27.3122	-1.8224	0.1385	-26.7762	-1.8520
	PSO	0.1466	-26.2160	-2.2012	0.1532	-24.9824	-2.2688
	LMS	0.2104	-21.3028	-1.7050	0.2307	-20.0240	-1.8846
FLANN	SOS	0.1134	-29.5440	-4.0114	0.1245	-26.9832	-3.5204
	CSO	0.1452	-27.8002	-2.4616	0.1604	-25.2014	-2.4022
	CLONAL	0.1296	-26.4226	-1.8054	0.1424	-24.6780	-1.8337
	PSO	0.1530	-26.1012	-2.0524	0.1614	-23.6120	-2.2304
	LMS	0.2354	-20.0822	-1.6218	0.2466	-19.6812	-1.6830
RBF	SOS	0.1202	-28.4245	-3.9126	0.1326	-26.5031	-3.4540
	CSO	0.1530	-27.3248	-2.4124	0.1649	-26.2310	-2.3986
	CLONAL	0.1322	-26.2120	-1.8210	0.1486	-25.2114	-1.8210
	PSO	0.1645	-25.8040	-2.0114	0.1682	-23.8220	-2.2210
	LMS	0.2540	-19.8825	-1.6100	0.2578	-19.1014	-1.5810
FIR	SOS	0.2566	-21.0422	-2.7145	0.2632	-22.0120	-2.6023
	CSO	0.2583	-20.1321	-2.2010	0.2614	-19.2014	-2.1986
	CLONAL	0.2710	-20.0124	-1.7132	0.2927	-18.8933	-1.7004

PSO	0.2654	-19.6482	-1.8240	0.2982	-19.3020	-1.7925
LMS	0.3022	-17.2125	-1.5232	0.3215	-17.1015	-1.5126

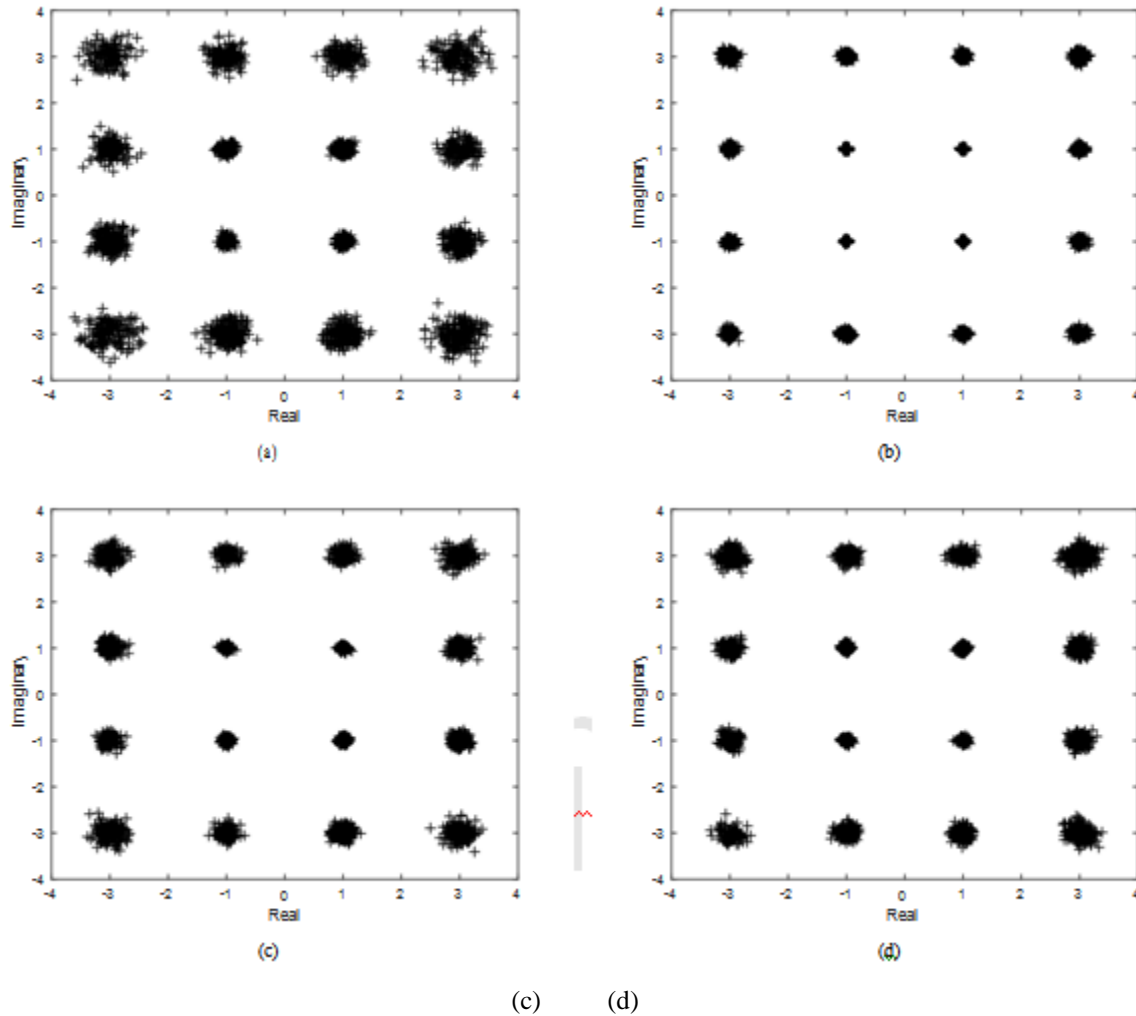


Figure 7: Constellation diagrams for telephone channel with 16-QAM : (a) Channel output without equalization (b) Performance after WNN-SOS equalizer (c) Performance after WNN-CSO equalizer (d) Performance after WNN-CLONAL equalizer.

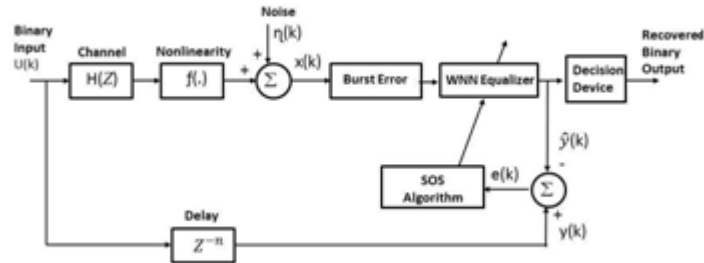


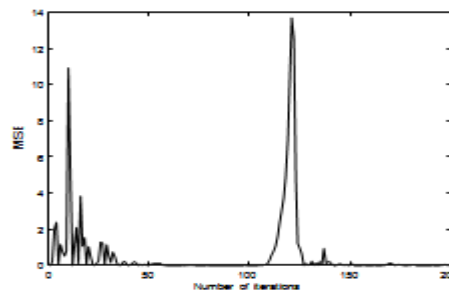
Figure 8: Block diagram of proposed WNN-SOS equalizer to handle burst error.

Table 3: Impulse response of telephone channel

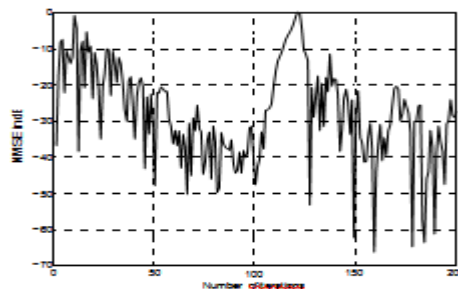
Tao No.	Real	Imaginary
0	0.0145	-0.0006
1	0.075	0.0176
2	0.3951	0.0033
3	0.7491	-0.1718
4	0.1959	0.0972
5	-0.2856	0.1896
6	0.0575	-0.2096
7	0.0655	0.1139
8	-0.0825	-0.0424
9	0.0623	0.0085
10	-0.0438	0.0034
11	0.0294	-0.0049
12	-0.0181	0.0032
13	0.0091	0.0003
14	-0.0038	-0.0023
15	0.0019	0.0027
16	-0.0018	-0.0014
17	0.0006	0.0003
18	0.0005	0.0000
19	-0.0008	-0.0001
20	0.0000	-0.0002
21	0.0001	0.0006

Table4:Comparative performance analysis of the proposed WNN SOS equalizer with other neural network and FIR based equalizers for Telecommunication channel 3

Channel 3		Training		Testing
Structure	Algorithm	MSE	NMSE (dB)	BER value 20 dB SNR
WNN	SOS	0.0812	-29.5432	-6.2320
	CSO	0.0945	-29.2120	-5.9824
	CLONAL	0.1084	-27.8912	-5.7432
	PSO	0.1220	-26.3430	-4.9120
	LMS	0.1448	-24.1058	-4.6214
FLANN	SOS	0.1124	-29.1014	-5.9040
	CSO	0.1231	-28.6218	-5.8025
	CLONAL	0.1288	-27.5238	-5.7202
	PSO	0.1482	-26.1226	-4.8024
	LMS	0.1513	-23.5014	-4.3212
RBF	SOS	0.1012	-29.4230	-6.0452
	CSO	0.1134	-28.3421	-5.9126
	CLONAL	0.1236	-26.6324	-5.8428
	PSO	0.1322	-26.0324	-4.9037
	LMS	0.1475	-23.8010	-4.5166
FIR	SOS	0.1535	-26.3050	-4.8020
	CSO	0.1642	-25.4510	-4.7125
	CLONAL	0.1688	-24.5046	-4.6830
	PSO	0.1715	-23.6628	-4.6021
	LMS	0.1820	-21.0213	-4.3033



(a)



(b)

Figure 9: Simulation study of of a traditional FIR equalizer trained by LMS algorithm to handle burst error : (a) MSE convergence plot of Channel 1 with non-linearity 1 and AWGN 30dB (b) NMSE plot.

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