RESEARCH ARTICLE

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A Noval method to improve the Logic Test by using Single Cycle Access Structure

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Abstract

This paper describes a new single cycle access test structure for logic testing in digital circuits. It avoids the peak power consumption problem of conventional shift based scan chains and reduces the activity during shift and capture cycles of the logic used. This leads to more realistic circuit behaviour during stuck-at and at-speed tests. It provides access to the complete test to run at much higher frequencies which is equal or close to the one in functional mode. The test cycles per net are based on a simple test pattern generator algorithm without using test pattern compression is below 1 for larger designs, and are independent of the design size. The structure allows an additional on-chip debugging signal visibility for each register in the scan chain. This method is backward compatible to full scan designs, existing test pattern generators and simulators can be used with a minor enhancement. It is shown, that how to combine the proposed solution with built-in self-test (BIST) and massive parallel scan chains.

I. Introduction

This paper presents a novel scan cell register for logic tests combined with a novel scan cell routing architecture. The structure allows a single cycle access (SCA) to individual register sets. This access scheme is fundamentally different to SS. It can be compared to a memory with single cycle synchronous write and asynchronous read functionality, whereas the remaining memory content (registers) does not change. The proposed structure is applicable for pattern driven tests and for BIST.

The paper provides reasonable data but is not limited to a frozen solution. It also describes about various trade-offs of different alternatives used in this BIST. The logic testing is a wide field and different users have different preferences on the applied and preferred logic. A reference example based on 992 registers is used.

The proposed FLS gating technique minimizes power in the combinational block. It removes redundant switching activity in the combinational block and also provides leakage minimization through application of the best input vector during scan shifting. The proposed scan partitioning technique reduces average and peak power in the scan chain by minimizing rippling of scan values deterministically.

II. Single Cycle Access Structure For Logic Test

SSR is a compression solution that does not require any information about don't care bits. Yet, it achieves 10 xs to 40x reduction in test data volume, test time, and test channel requirements. With the same minimal hardware used in as SAS, SSR achieves this major cost reduction through modifying

the ATPG process instead of utilizing the don't care bits.

This scheme where the scan enables control information for the launch and capture cycle is already embedded itself in the test data. A new scan register, called the last transition generator (LTG), generates the local fast scan enable signals. The LTG cell has the flexibility to be inserted anywhere in the scan chain and the hardware area overhead is comparable to the pipeline scan enable approach. The proposed method poses no additional constraints for the place and route tool and provides more flexibility to re-order the scan cells to meet the timing closure of the local scan enable signals.

Our goal is to reduce all the VPT parts simultaneously. This is achieved by reducing test data Volume with variable-to-fixed run-length coding and reducing scan-in power dissipation as test application time with RAS structure. A heuristic algorithm is proposed to arrange the scan cells in RAS to obtain a high compression ratio.

The peak power consumption problem of conventional shift-based scan chains and reduces the activity during shift and capture cycles. This raises to more realistic circuit behaviour during stuck at faults and at-speed logic tests. It enables to complete test run at very higher frequencies equal or close to the one in functional mode. It will be shows, that a lesser number of test cycles can be achieved as compared to other solutions. The test cycle per scan chain based on a simple test pattern generator algorithm without Sri Lakshmi. Chandana et al Int. Journal of Engineering Research and Applications www.ijera.com ISSN: 2248-9622, Vol. 3, Issue 6, Nov-Dec 2013, pp.1721-1726

test pattern compression is below 1 for larger designs and is independent of the scan chain size. The structure provides an additional on-chip debugging signal visibility for each register in scan chain. This method is backward compatible to full scan designs and existing test pattern generators and simulators can be used with a minor enhancement. It shows how to combine the proposed solution with built-in self test (BIST) and massive parallel scan chains.

2.1 SCAh-FF based on an S-FF:



- The key element of the single cycle access structure with hold mode is the signal cycle access register with hold mode (SCAh-FF).
- It is based on a standard scan register (S-FF) and uses two more 2-to-1 multiplexers. The new SCAh-FF register can be seen in Fig.



2.2 SCAH-FF Connectivity:

Fig. 2 shows the SCAh-FF and its connectivity. The two major differences are, that the scan-in $\{si \}$ is now connected to a dedicated scanout $\{so \}$ of the preceding register in the scan chain and the register $\{se[1]\}$ inputs on the same scan depth are connected to the same line-select $\{ls \}$ signal, which is driven by a "1 to *N*" decoder. Single cycle accesses withhold mode (SCAh-ff) register connected to the line-select signal and it is considered as one line. If {add} is 0, no line is selected. {se[0]} of each SCAh-FF is connected to the global scan enable signal { gse} (comparable to the global scan enable signal of shift-scan structures). The output of the address decoder is connected to the {se[1] } pin of the registers on same line select signal on a particular line. All registers on the same scan depth, enabled by the same line-select signal can be read or written with a single cycle access.





Fig – 3: SCA-Structure without Hold mode

In order to reduce the area overhead of a single cycle access with hold mode, a simpler Single Cycle Access (SCA-FF) is discussed. It adds only one 2x1 Multiplexer to the standard S-FF. It only has one {se} input, which is connected to the individual line-select signal (ls), which is connected to the global enable signal in the Single Cycle Access with hold mode register is removed, therefore the entire global {se} tree is incomplete. The SCA-structure (SCAS) connectivity and page organization equals the one of the SCAhS without the global scan enable {gse}.

2.4 Gated SCA-Structure:

The hold function of the SCAh-FF is missing in the Single cycle access register; it is instead built into the gated clock tree of the gSCAS. Fig shows the connectivity of the gSCA. The scan path reaches from the scan-in AND-selector over the SCA-FF chain (by connecting the scan-out pins of each SCA-FF with the scan-in pins of the succeeding SCA-FF) and is connected with the input of the XOR-tree. The individual line-select signals {ls} are connected with the {se} input of the SCA-FF in the same line. All Single Cycle Access registers on a line are clocked by a gated clock element (gcl). The gcl is driven by the clock and the line-select signal. The gcl can be enhanced, if a clock enable signal {ce} generated by combinatorial logic exists. The gscan enable signal is connected with each {gcl}, which is already the case in SS if gated clock elements are used to propagate the clock during shift.

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2.5 SCAH-FF Page:

The SCAhS enables single cycle read/write accesses to the each register line. The test structure is organized as pages, to achieve a read/write access at design speed or at a reasonable test speed.

The page depth equals the scan chain depth (SD = number of SCAh-FF connected to one chain on one page). Assuming it is 31. Multiplied with the scan width (SW = number of scan chains on one page, for example 32), then number of SCAh-FF is SD*SW=992 per page.

III. Implementation of the modules:

All the sub modules were implemented and simulated with Xilinx 13.2 version based on Spartan 3 Basic kit hardware specifications. Initially all modules implemented separately and then synthesized in our proposed design implementation as given below.

Various implementations with and without hold mode as well as gated and partial implementation methods are presented. The aspects feasibility, peak power consumption, switching activity during test, area, test cycles, at-speed testing and debugging features are compared. The best solution (gSCAS) is compared to RAS implementations and is superior to all known RAS solutions







1 OUT OF N DECODER



2*1 MUX:

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SCAN FLIPFLOP



BENCHMARK :

Messages		
♦ /VS27Combinational_logic/G0	St1	
VS27Combinational_logic/G1	St0	
/VS27Combinational_logic/G2	St1	
/VS27Combinational_logic/G3	St1	
/VS27Combinational_logic/G5	St0	
/VS27Combinational_logic/G6	St1	
/VS27Combinational_logic/G7	St0	
/VS27Combinational_logic/G20	St1	
💠 /VS27Combinational_logic/G19	St1	
💠 /VS27Combinational_logic/G17	St0	
💠 /VS27Combinational_logic/G10	St0	
💠 /VS27Combinational_logic/G11	St1	
ARTTCombinational India/C12	C+0	
A 📰 💿 🛛 Now	200 ns	is 200 ns
🔒 🆉 😔 Cursor 1	0 ns	0 ns

SCAH FLIPFLOP



SCAHS CONNECTIVITY



SCAH-FF PAGE

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St1	
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St1	
St0	
St0	
Now 1600 ns 000 as 000 as 000 as	
	11130000000000000000000000000000000000





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TOP MODULE:



DEVICE UTILIZATION SUMMARY

Device Utilization Summary Logic Utilization Used Available Utilization Note(s Number of Slice Flip Flops 7,168 1% 8 Number of 4 input LUTs 1% 27 7 168 Logic Distribution 16 3,584 1% Number of occupied Slices Number of Slices containing only related logic 16 16 100% 0 16 ٥% Number of Slices containing unrelated logic 28 7,168 1% Total Number of 4 input LUTs Number used as logic 27 Number used as a route-thru 1 17 141 Number of bonded IOBs 12% IOB Flip Flops 2 1 8 12% Number of GCLKs Total equivalent gate count for design 257 Additional JTAG gate count for IOBs 816

RTL SCHEMATIC



TECHNOLOGY SCHEMATIC



IV. CONCLUSION

A single cycle access structure is explained. Various blocks implementations with and without mode as well as gated and partial hold implementation methods are presented. The aspects feasibility, peak power consumption, switching activity during test, area, test cycles, at-speed testing and debugging are compared. A guide is given how to choose the best implementation. The best solution (gSCAS) is compared to RAS implementations and is superior to all known RAS solutions. If BIST is preferable due to limited chip IOs or partial scan implementation, an address controlled BIST is discussed. The ATPG algorithms can be enhanced with the same methods SS implementations are optimized. Future work is related to algorithms for reducing the test cycles per net itself, register reordering, and pattern optimization for activity reduction and de-compression methods for BIST using the gSCAS.

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