

Figure 1. Telescopic OTA

input and output swing. In order to alleviate some of the drawbacks of telescopic operational amplifier, a folded cascode OTA based on Wilson mirror can be used.

II.I. Basic configuration CMOS Folded Cascode OTA

The operational transconductance amplifier (OTA) is used as basic building block in many switched capacitor filters. OTA is basically an op-amp without an output buffer and can only drive capacitive loads [7], [8].

An OTA is an amplifier where all nodes are low impedance except the input and output nodes. A useful feature of OTA is that its transconductance can be adjusted by the bias current. Filters made using the OTA can be tuned by changing the bias current I_{bias} [9]. Two practical concerns when designing an OTA for filter applications are the input signal amplitude and the parasitic input/output capacitances.

Large signals cause the OTA gain to become non-linear. The external capacitance should be large compared to the input or output parasitic of the OTA. This limits the maximum frequency of a filter built with an OTA and causes amplitude or phase errors. These errors can usually be reduced with proper selection of I_{bias} . The performance of simple OTA is limited by its input and output voltage swing. To overcome these limits of simple OTA and have an improved performance a Folded Cascode OTA is used.

The folded cascode OTA is shown in Fig. 2. The name "folded cascode" comes from folding down n-channel cascode active loads of a diff-pair and changing the MOSFETs to p-channels. Folded cascode OTA has a differential stage consisting of PMOS transistors M9 and M10 intend to charge Wilson mirror. MOSFETs M11 and M12 provide the DC bias voltages to M5-M6-M7-M8 transistors. [10] Apply AC input Voltage between V_+ and V_- , cause the diff-amplifier drain current to become $g_m V_{in}$. This AC differential drain current is mirrored in the cascaded MOSFETs M1 to M6.

The output Voltage of the OTA is given by:

$$V_{out} = G_m V_{in} R_o \quad (1)$$

The "Unity gain frequency" of the OTA is:

$$F_u = 2 \pi g_m / C_L \quad (2)$$

G_m is computed as:

$$G_m = 2 \pi GBW C_L \quad (3)$$

The open-loop voltage gain is given by:

$$A_v = \frac{\{g_{m9} g_{m4} g_{m6}\}}{I_D^2 (g_{m4} \lambda_N^2 + g_{m6} \lambda_P^2)} \quad (4)$$

Where g_{m9} , g_{m4} and g_{m6} are respectively the transconductances of transistors M9, M4 and M6. I_D is the bias current flowing in MOSFETs M4, M6, and M9. Like, C_L is the capacitance at the output node. λ_N and λ_P are the

parameters related to channel length modulation respectively for NMOS and PMOS devices. Taking the complementarities between the Transistors M4 and M6 into account:

$$g_{m4} = g_{m6} \quad (5)$$

The gain expression becomes:

$$A_v = \frac{\{g_{m9} g_{m4}\}}{I_D^2 (\lambda_N^2 + \lambda_P^2)} \quad (6)$$

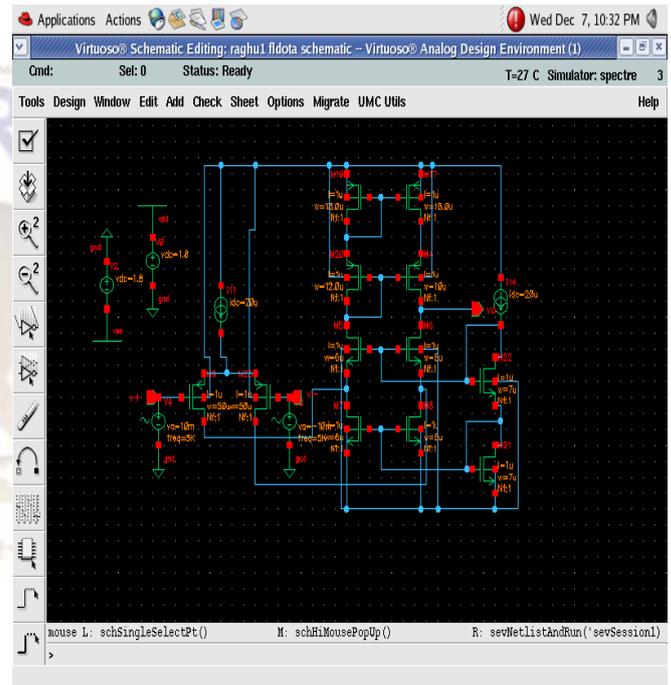


Figure 2. Folded Cascode OTA

III. FOLDED CASCODE OTA DESIGN METHODOLOGY.

To show folded cascode OTS performances, this paper is interested in OTA design carrying. This design follows synthesis procedure based on the g_m/I_D methodology[11].

III.1. Sizing Algorithm

MOS transistors are either in strong inversion or in weak inversion. The design methodology based G_m/I_D characteristic, proposed by allows a unified synthesis methodology in all regions of operation the MOS transistor. We consider the relationship between the ratio of the transconductance G_m over the DC drain current I_D , and the normalized drain current $I_D / (W/L)$ as a fundamental design relation[7]. G_m/I_D are based on its relevance for the following reasons:

- It is strongly related to the performance of analog circuits;
- It gives an indication of the device operation Region;
- It provides a simple way to determine the transistor dimensions.

$$F_u \rightarrow 2 \pi g_{m9} / I_D \leftrightarrow \frac{I_D}{(W/L)_9}$$

$$A_v \rightarrow g_{m4} / I_D \leftrightarrow \frac{I_D}{(W/L)}$$

Figure 3. Design Plan

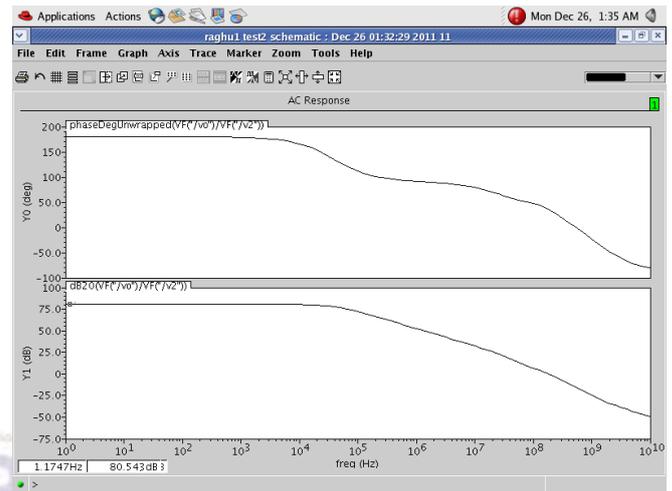


Figure 6. Gain and Phase Curve

III.2. OTA Design

After applying the design strategy, we obtained the parameters computed and summarized in Table 1.

Transistor	W(μm)	L(μm)
M ₁	18	1
M ₂	18	1
M ₃	12	1
M ₄	10	1
M ₅	6	1
M ₆	6	1
M ₇	6	1
M ₈	6	1
M ₉	50	1
M ₁₀	50	1
M ₁₁	7	1
M ₁₂	7	1

Table 1. Width and lengths of different transistors

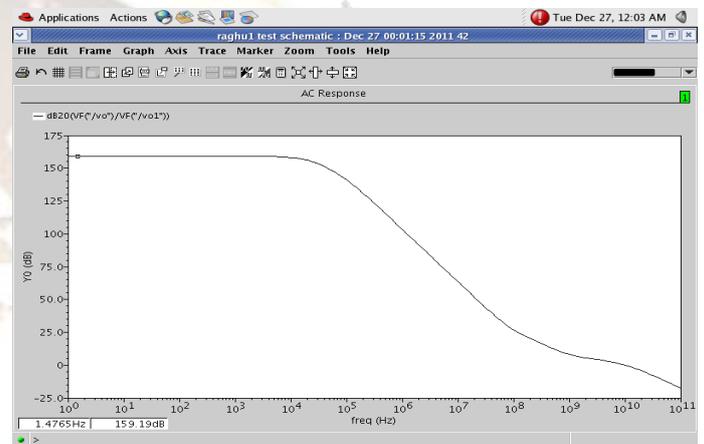


Figure 8. CMRR

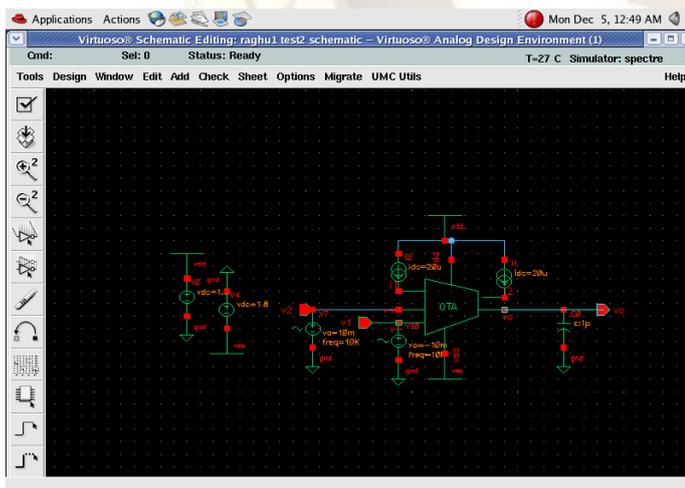


Figure 6. Test Bench of Folded Cascode OTA

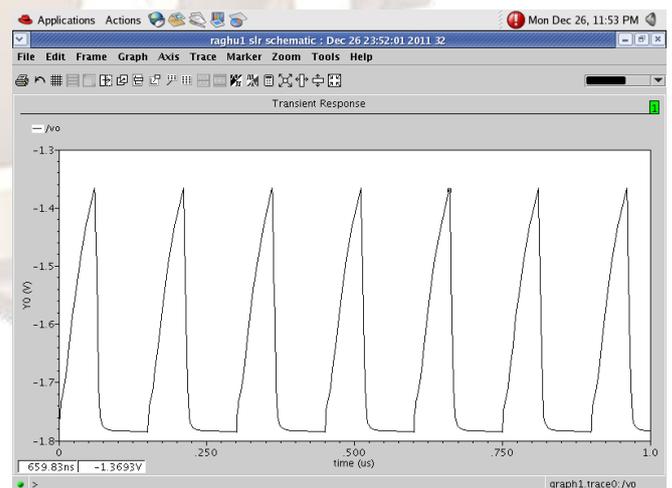


Figure 9. Slew Rate

III.3. Results

The designed Folded Cascode OTA was biased at 1.8V power supply voltage using CMOS technology of 0.18μm with the BSIM3V3 MOSFET model.

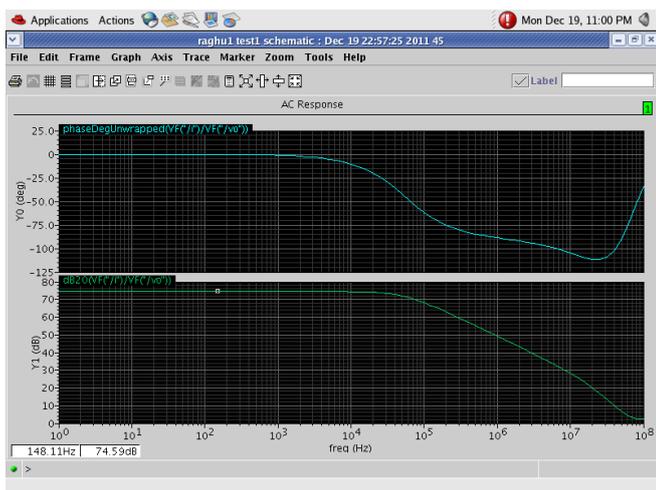


Figure 10. PSRR

Results summary

Specifications	Results
Gain	80.219 dB
GBW	452 MHz
3dB bandwidth	42.7 KHz
F_T	176 MHz
Phase Margin	40 Degree
Gain Margin	15dB
Load	0.1pf
Technology	0.18 μ m
Supply Voltage	$\pm 1.8V$
Power Dissipation	126 μ W
Slewrate	5 V/usec
CMRR	159.19
PSRR	74.59

IV CONCLUSIONS

Since the Folded Cascode OTA based on Wilson mirror has a limited output swing. For the folded Cascode OTA using a Wilson mirror, the maximum output voltage is set lower than: $V_{dd} + V_T + 2V_{dsat}$, so, we can use cascode mirror to compensate the fall to $+2V_{dsat}$. This paper presents an efficient OTA design, so, the goal to reach moderate gain and large bandwidth. Transconductance cells are relatively simple circuits which allow operating for high frequencies. Future work involves the search of low power consumption and Ultra low-supply voltage structure, an update to nanotechnology process for RF application.

REFERENCES

[1] M.G.R. Degrauwe et al., "IDAC: An interactive design tool for analog CMOS circuits," IEEE J.Solid-State circuits, vol. sc-22, no. 6, dec.(1987), pp. 1106-1116.

[2] R.Harjani, R.A. Rutenbar and L.R. Carley, "OASYS: A framework for analog circuit synthesis," IEEE Trans. Computer-Aided Design, vol. 8, no. 12, Dec. (1989), pp. 1247-1266.

[3] H. Y. Koh, C.H Séquin and P.R. Gray, "OPASYN: A compiler for CMOS operational amplifiers," IEEE Trans. Computer-Aided Design, vol. 8, no. 12, Dec. (1990), pp. 113-125.

[4] J.P. Harvey, M.I. Elmasry and B. Leung, "STAIC: An interactive framework for synthesizing CMOS band BiCMOS analog circuits," IEEE Trans. Computer-Aided Design, vol. 11, no. 11, Nov. (1992), pp. 1402-1417.

[5] M. Fakhfakh, M. Loulou, and N. Masmoudi, "Optimizing performances of switched current memory cells through a heuristic," Journal of Analog Integrated Circuits and Signal Processing, Springer Editor, (2006).

[6] M.hershenson, S.Boyd, and T. Lee. "Optimal design of a CMOS op-amp via geometric programming". Stanford.edu/prople/boyd

[7] F. Silveira, D. Flandre et P.G.A. Jespers, "A gm/ID based methodology for the design of CMOS analog circuits and its application to the synthesis of a SOI micropower OTA", IEEE J. of Solid State Circuits, vol. 31, n. 9, sept. 1996.

[8] M. Banu, J. M. Khoury, and Y. Tsvividis, "Fully Differential Operational Amplifier with Accurate Output Balancing," IEEE Journal of Solid State circuits, Vol. 23, No. 6, pp. December 1990.

[9] R. Hogervorst, J. P. Tero, R. G. H. Eschauzier, and J. H.Huijsing, "A Compact Power efficient 3 V CMOS Rail-to-Rail Input/Output Operational Amplifier for VLSI cell Libraries," IEEE Journal of Solid State Circuits, Vol. 29,pp. December 1988.

[10] Houda Daoud, Samir Ben Salem, Sonia Zouari, Mourad Loulou, "Folded Cascode OTA Design for Wide Band Applications", Design and Test of Integrated Systems in Nanoscale Technology, 2006.

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